

Fast Polar Fourier Transform

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Scientific Computing and Computational Mathematics

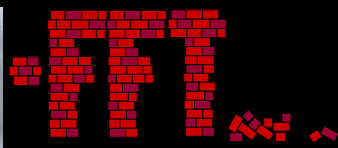
Stanford University

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Image and Signal Processing Workshop

IMA - Minneapolis

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Amir Averbuch (TAU-CS-Israel), and Ronald Coifman (Yale-Math)



Collaborators



Dave Donoho

Statistics Department
Stanford



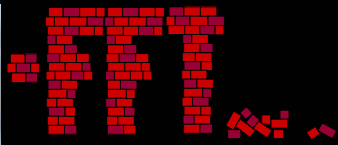
Amir Averbuch

CS Department
Tel-Aviv University



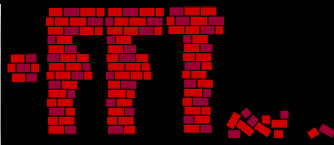
Ronald Coifman

Math. Department
Yale




Fast Polar Fourier Transform

- ❑ FFT is one of top 10 algorithms of 20th century.
- ❑ We'll extend utility of FFT algorithms to new class of settings in image processing.
- ❑ Create a tool which is:
 - Free of emotional involvement, &
 - Freely available over the internet.
- ❑ Current Stage:
 - We have the tool, and its analysis,
 - Have not demonstrated applications yet.

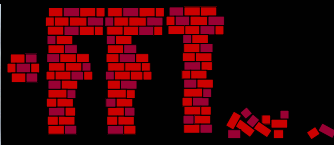


Agenda

1.  Thinking Polar – Continuum
2. Thinking Polar – Discrete
3. Current State-Of-The-Art
4. Our Approach - General
5. The Pseudo-Polar Fast Transform
6. From Pseudo-Polar to Polar
7. Algorithm Analysis
8. Conclusions

Background
&
Motivation

New
Approach
and its
Results



1. Thinking Polar - Continuum

□ For today $f(x,y)$ function of $(x,y) \in \mathbb{R}^2$

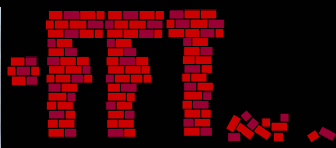
□ Continuous Fourier Transform

$$\hat{f}(u, v) = (\mathfrak{F}f)(x, y) = \int \int f(x, y) \exp\{-ixu - iyv\} dx dy$$

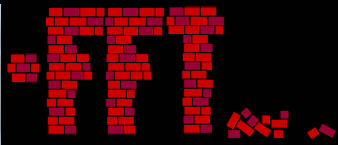
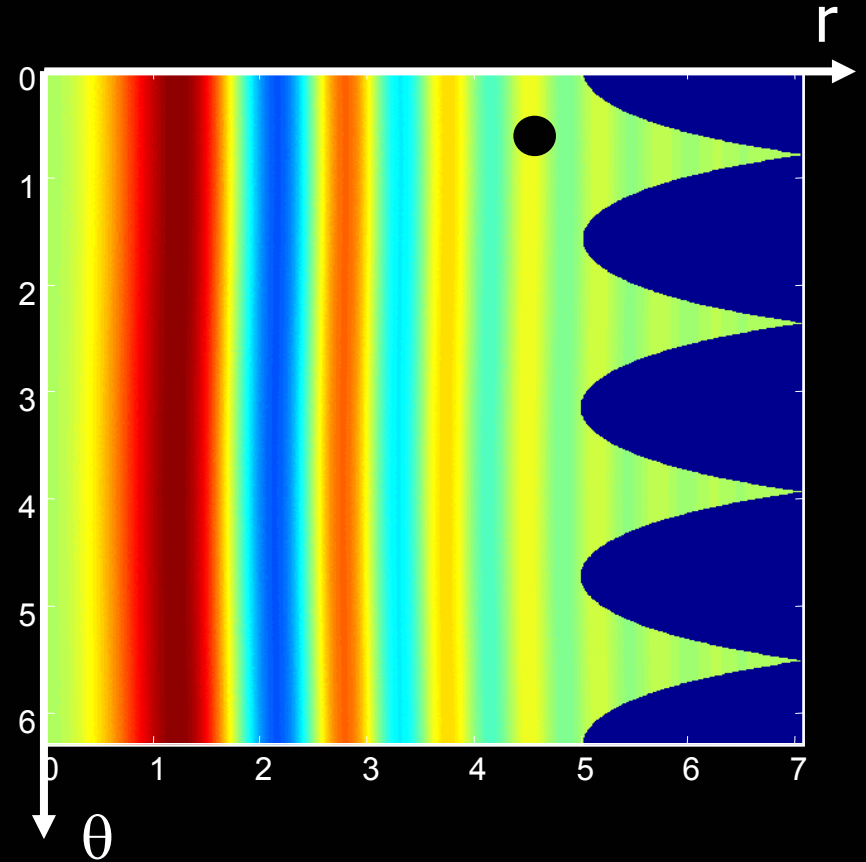
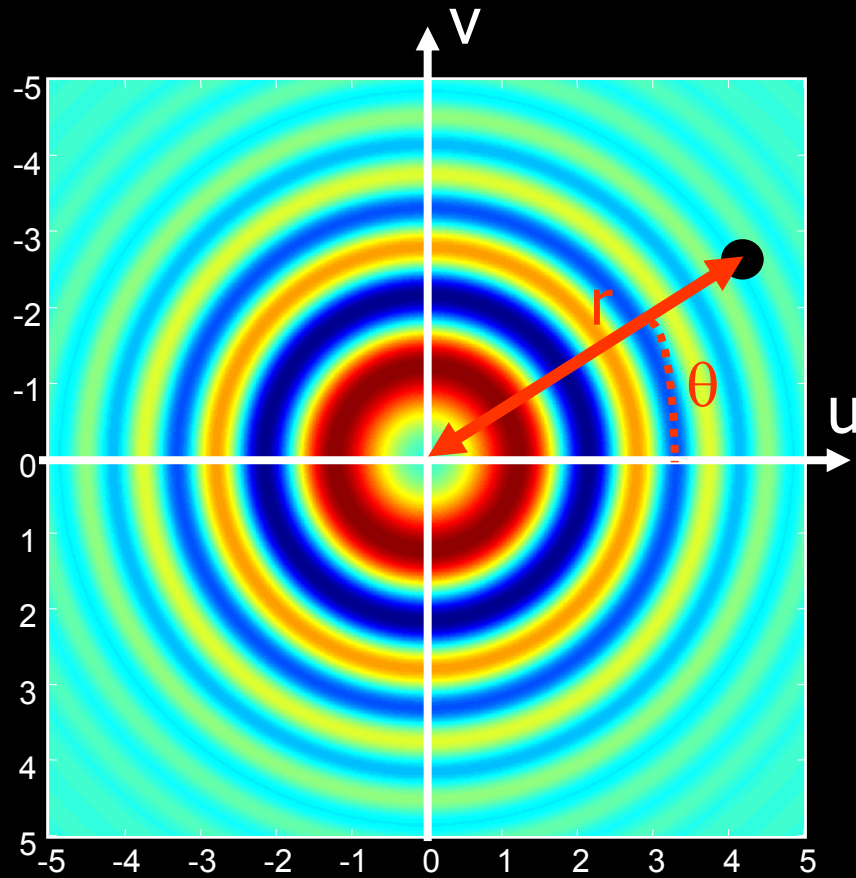
□ Polar coordinates: $u=r \cdot \cos(\theta)$, $v=r \cdot \sin(\theta)$

$$\begin{aligned} \tilde{f}(r, \theta) &= \hat{f}(r \cdot \cos(\theta), r \cdot \sin(\theta)) = \\ &= \int \int f(x, y) \exp\{-ixr \cdot \cos(\theta) - iy \cdot \sin(\theta)\} dx dy \end{aligned}$$

□ Important Processes easy to continuum polar domain.



1. Thinking Polar - Continuum



Natural Operations: 1. Rotation

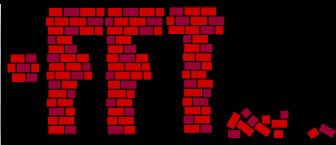
Using the polar coordinates, rotation is simply a shift in the angular variable.

□ Q_{θ_0} – planar rotation by θ_0 degrees

□ Rotation $f_{\theta_0}(x, y) = f(Q_{\theta_0}\{x, y\})$

□ In polar coordinates – shift in angular variable

$$\tilde{f}_{\theta_0}(r, \theta) = \tilde{f}(r, \theta - \theta_0)$$



Natural Operations: 2. Scaling

Using the polar coordinates, 2D scaling is simply a 1D scaling in the radial variable.

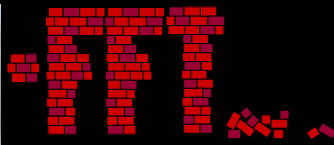
□ S_α – planar scaling by a factor α

□ Scaling $f_\alpha(x, y) = f(S_\alpha\{x, y\})$

□ In polar coordinates – 1D scale in radial variable

$$\tilde{f}_\alpha(r, \theta) = \text{Const} \cdot \tilde{f}(\alpha r, \theta)$$

□ Log-Polar – shift in the radial variable.



Natural Operations: 3. Registration

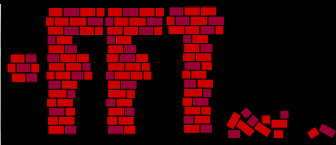
Using the polar coordinates, rotation+shift registration simply amounts to correlations.

□ $f(x,y)$ and $g(x,y)$: $f(x,y) = g(Q_{\theta_0} \{x,y\} + \{x_0, y_0\})$

□ Goal: recover $\{x_0, y_0, \theta_0\}$.

□ Angular cross-correlation between $|\tilde{f}(r, \theta)|$ and $|\tilde{g}(r, \theta)|$
– Estimate θ_0 .

□ Rotation & cross-correlation on regular Fourier transform gives the shift.



Natural Operations: 4. Tomography

Using the polar coordinates, we obtain a method to obtain the Inverse Radon Transform.

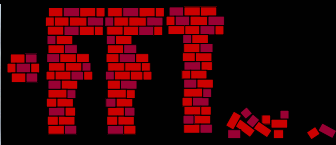
□ Radon Transform:

$$Rf(t, \theta) = \iint f(x, y) \delta(x \cos(\theta) + y \sin(\theta) - t) dx dy$$

□ Goal: Given $Rf(t, \theta)$, recover f .

□ Projection-Slice-Theorem: $(\mathfrak{T}_1 Rf)(t, \theta) = \tilde{f}(r, \theta)$.

□ Reconstruction: $Rf \mapsto \tilde{f} \mapsto \hat{f} \mapsto f$.



More Natural Operations

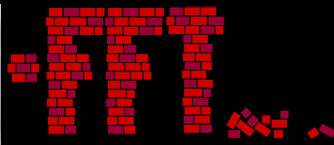
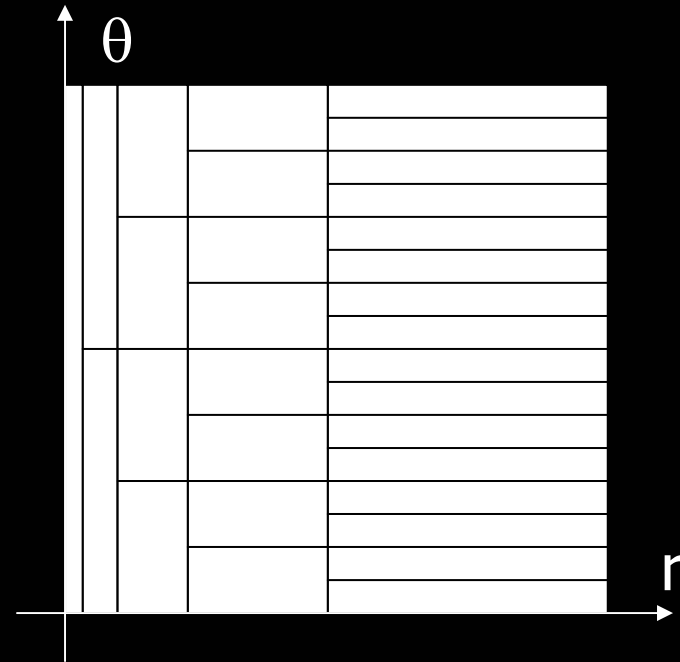
□ New orthonormal bases:

- Ridgelets,
- Curvelets,
- Fourier Integral operations,
- Ridgelet packets.


□ Analysis of textures.

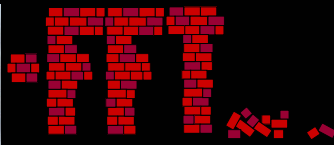
□ Analysis of singularities.

□ More ...



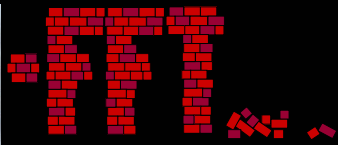
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2. Thinking Polar - Discrete

- Certain procedures very important to digitize
 - Rotation,
 - Scaling,
 - Registration,
 - Tomography, and
 - ...
- These look so easy in continuous theory – Can't we use it in the discrete domain?
- Why not Polar-FFT?



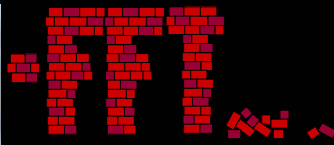
The FFT Miracles

□ 1D Discrete Fourier Transform

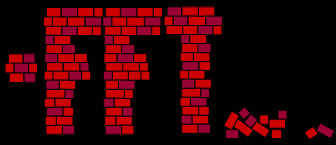
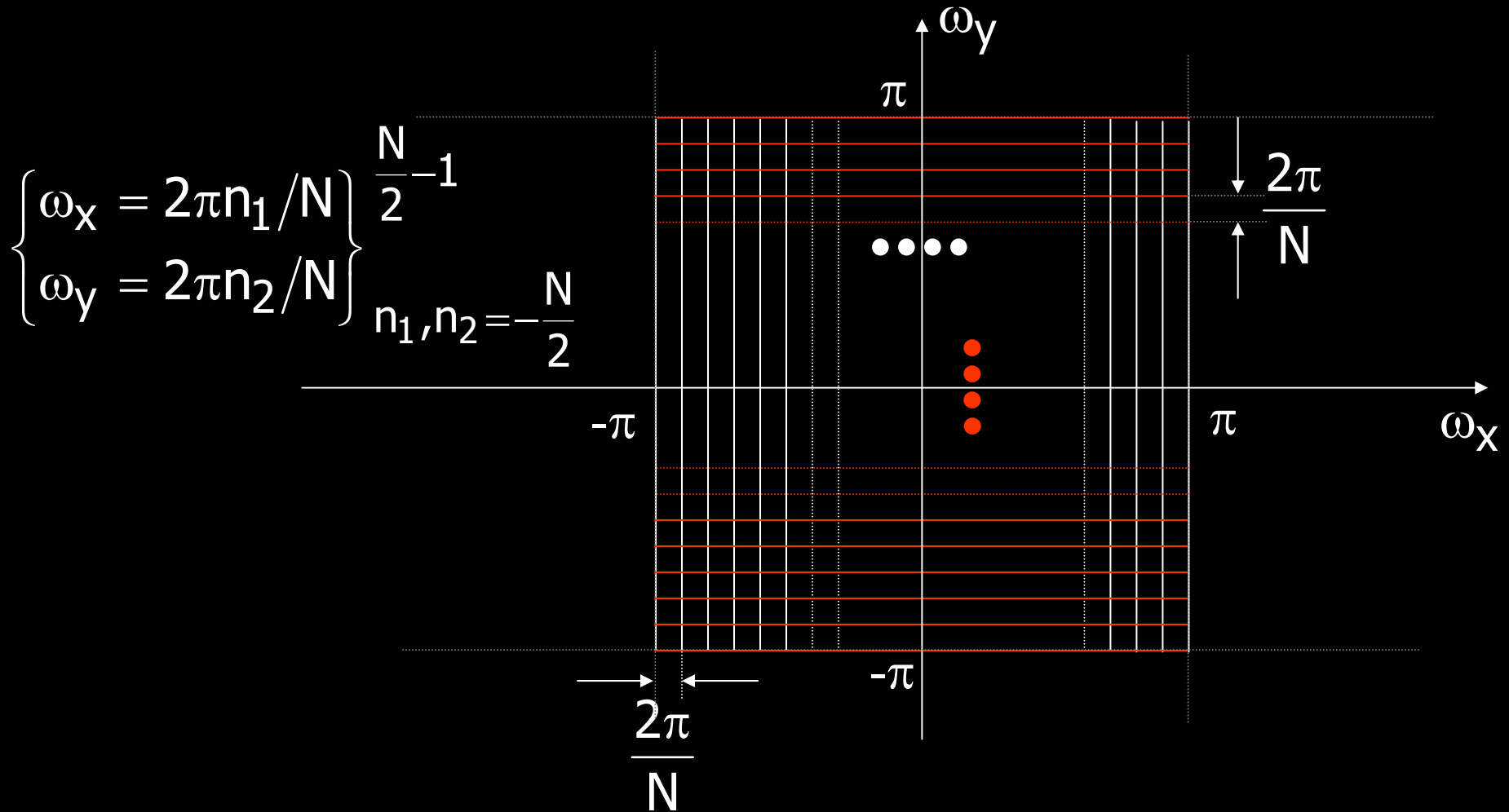
- Uniformly sampled in time and frequency – FFT.
- Complexity – $O(5N\log_2N)$ instead of $O(N^2)$.

□ 2D Discrete Fourier Transform

- Cartesian grid in space and frequency – Separability
- Only 1D-FFT operations.
- Smart memory management.

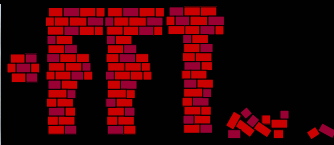
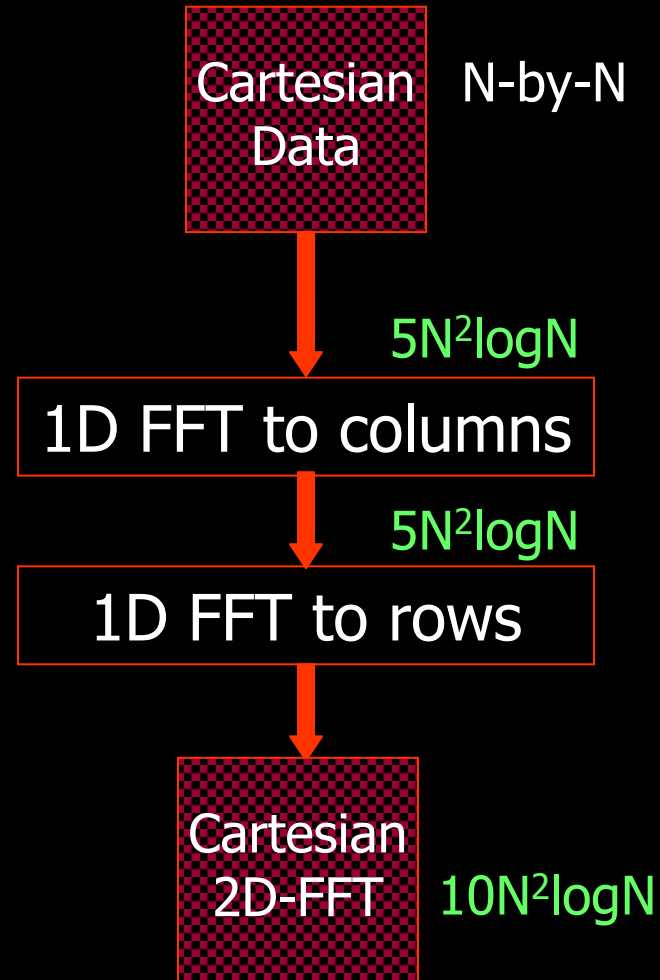


2D DFT – Cartesian Grid



2D FFT Complexity

- ❑ Complexity: $O(10N^2\log_2N)$ instead of $O(N^4)$.
- ❑ Important Feature: All operations are 1D
– leading to **efficient cache usage**



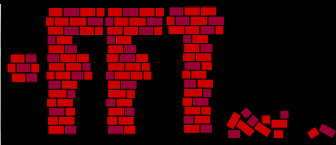
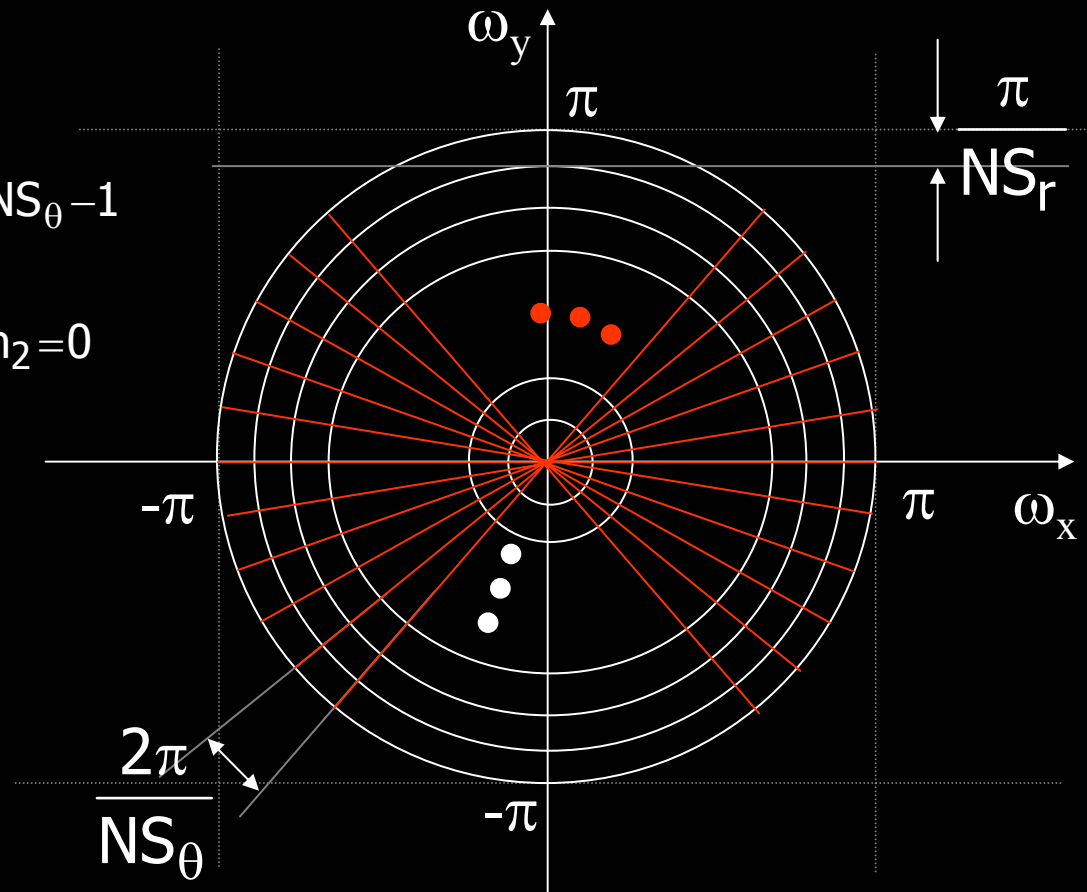
Discrete Polar Coordinates?

Choice of grid?

$$\left\{ r = \frac{\pi n_1}{NS_r} \right\}_{n_1=0}^{NS_r-1}, \left\{ \theta = \frac{2\pi n_2}{NS_\theta} \right\}_{n_2=0}^{NS_\theta-1}$$

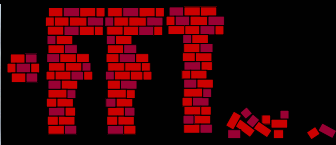
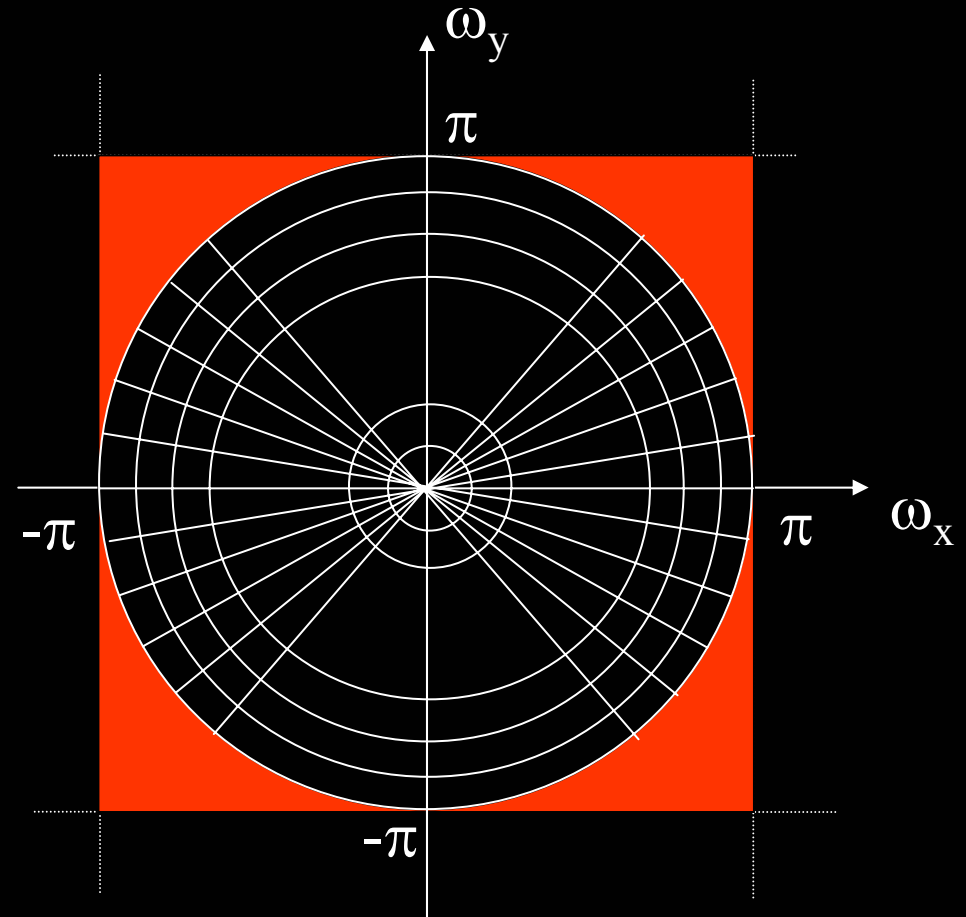
Resulting with NS_θ
rays with NS_r
elements on each:

For $S_\theta = S_r = 1$, we
have N^2 grid points.



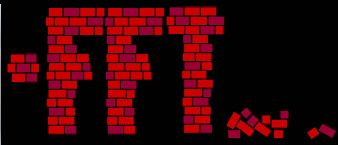
Grid Problematics

- ❑ Grid spacing?
- ❑ Fate of corners?
- ❑ No X-Y separability !!




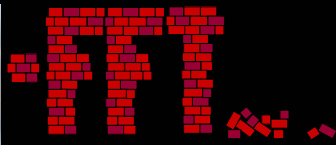
Polar FFT - Current Status

- ❑ Current widespread belief - There cannot be a fast method for DFT on the polar grid. See e.g. **The DFT: an owner's manual**, Briggs and Henson, SIAM, 1995, p. 284.
- ❑ Consequence of Non-existence:
 - Continuous Fourier – vague inspiration only.
 - Fourier in implementations widely deprecated.
 - Fragmentation: each field special algorithm.



Agenda

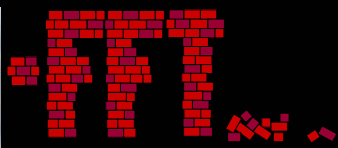
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3. Current State-Of-The-Art

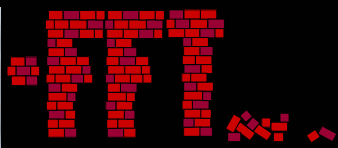
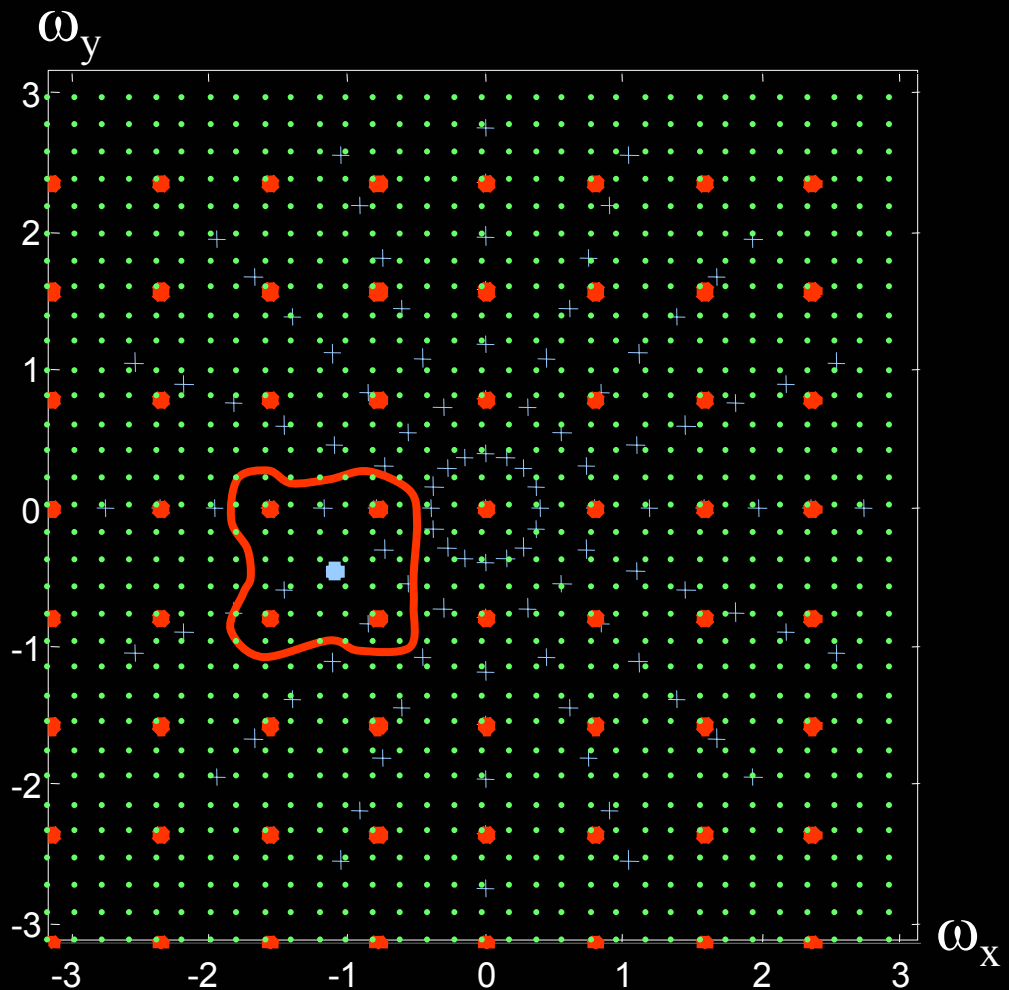
- Assessing T : Unequally-spaced FFT (USFFT)
 - Data in Cartesian set.
 - Approximate transform in non-Cartesian set.
 - Oriented to 1D – not 2D and definitely not Polar.

- Assessing T^\dagger : For Tomography
 - Data in Polar coordinates in frequency.
 - Approximate inverse transform to Cartesian grid.
 - Inspired by the projection-slice-theorem.



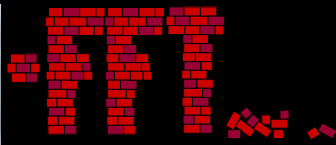
USFFT - Rational

- + Destination Polar grid
- Critically sampled Cartesian grid
- Over-sampled Cartesian grid



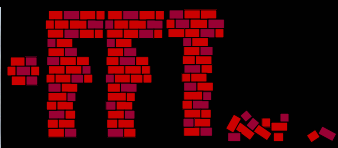
USFFT - Detailed

- ❑ **Over-sample** Cartesian grid.
- ❑ Rapidly evaluate FT:
 - Values F .
 - Possibly - partial derivatives.
- ❑ Associate Cartesian neighbors to each polar grid point.
- ❑ Approximate **interpolation**.



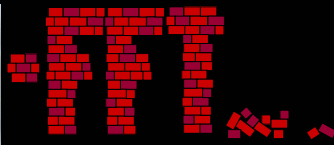
Our Reading of Literature

- Boyd (1992) – **Over-sampling** and **interpolation** by Euler sum or Lagrangian interpolation.
- Dutt-Rokhlin (1993,1995) - **Over-sampling** and **interpolation** by the Fast-Multipole method.
- Anderson-Dahleh (1996) – **Over-sampling** and obtaining the partial derivatives, and then **interpolating** by Taylor series.
- Ware (1998) – Survey on USFFT methods.




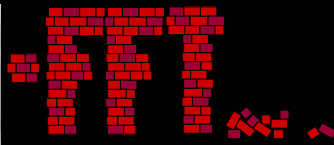
USFFT Problematics

- ❑ Several involved parameters:
 - Over-sampling factor,
 - Method of interpolation, and
 - Order of interpolation.
- ❑ Good accuracy calls for extensive over-sampling.
- ❑ Correspondence overhead: spoils vectorizability of algorithm - causes high cache misses.
- ❑ Emotionally involved.



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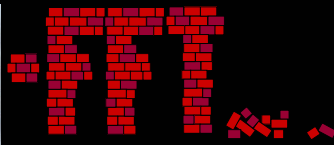
4. Our Approach - General

We propose a

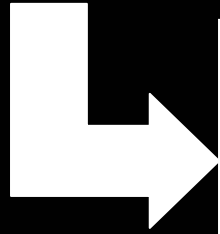
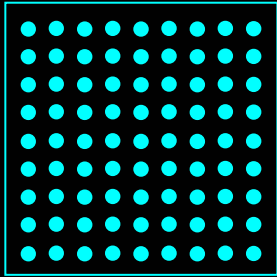
Fast Polar Fourier Transform

with the following features:

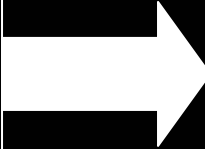
- Low complexity – $O(\text{Const} \cdot N^2 \log_2 N)$
- Vectorizability – 1D operations only
- Non-Expansiveness – Factor 2 (or 4) only
- Stability – via Regularization
- Accuracy – 2 orders of magnitude over USFFT methods



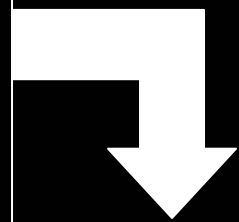
Our Strategy



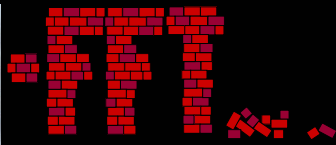
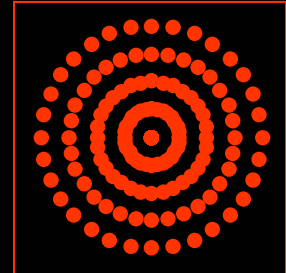
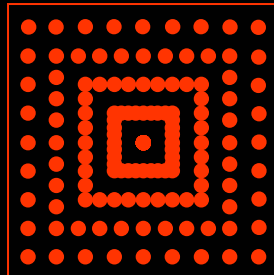
Fast and Exact
Fourier Trans.
on a polar-like
grid




1D
interpolations
to the polar
grid

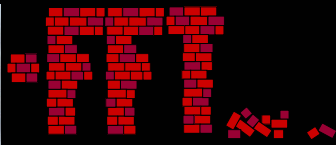


Pseudo
Polar
Grid



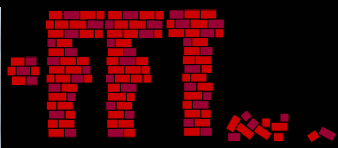
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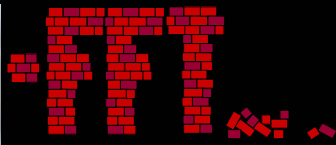
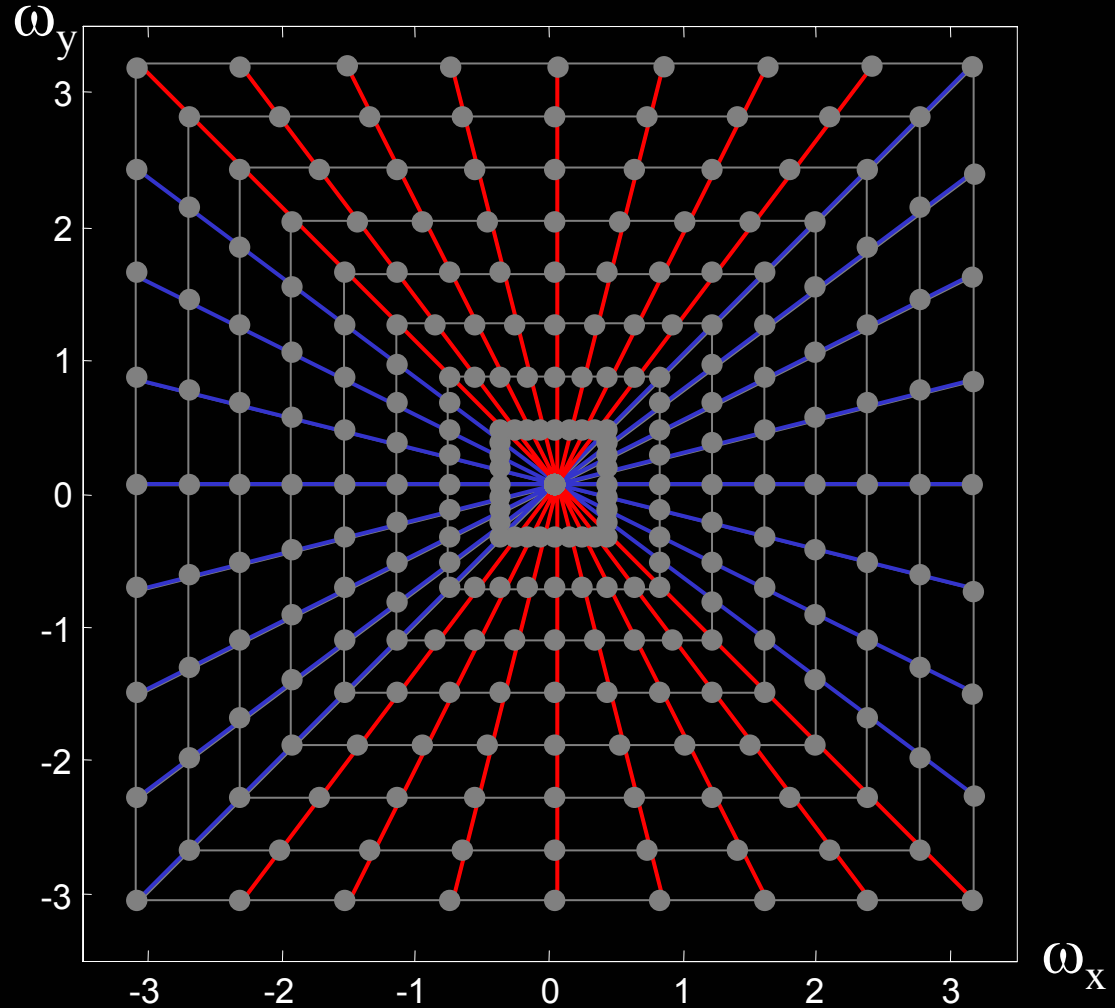
5. The Pseudo-Polar FFT

- Developed by Averbuch, Coifman, Donoho, Israeli, and Waldén (1998).
- Basic idea: A “Polar-Like” grid that enables
 - EXACT Fourier transform,
 - FAST computation,
 - 1D operations only.
- Applications: Tomography, image processing, Ridgelets, and more.



The Pseudo-Polar Skeleton

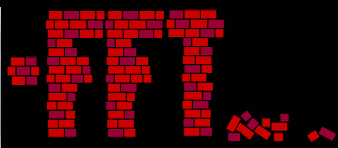
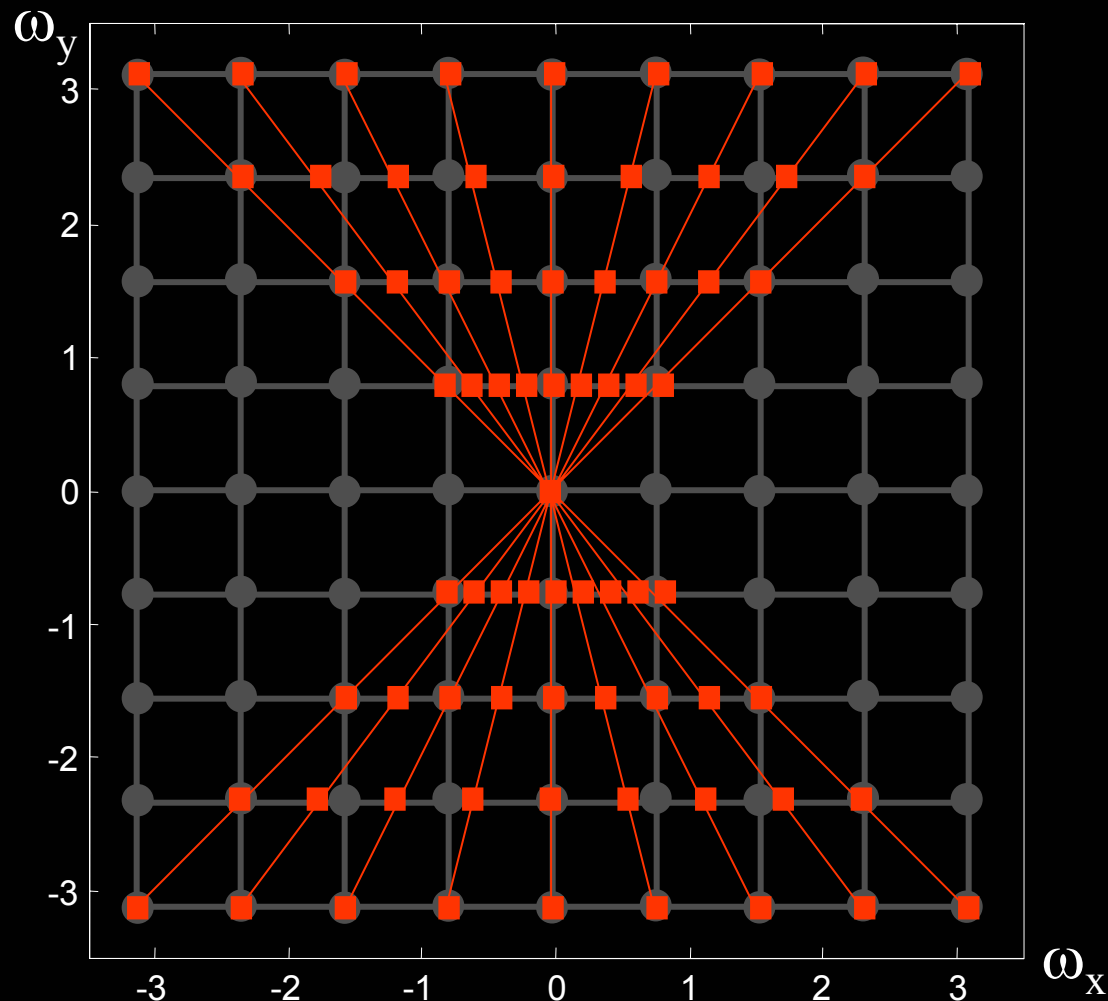
- NS_r equi-spaced concentric squares,
- NS_t 'equi-spaced' (not in angle)
- We separate our treatment to basically vertical and basically horizontal lines.



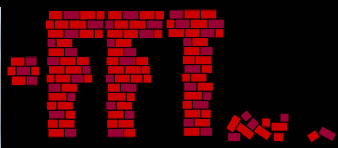
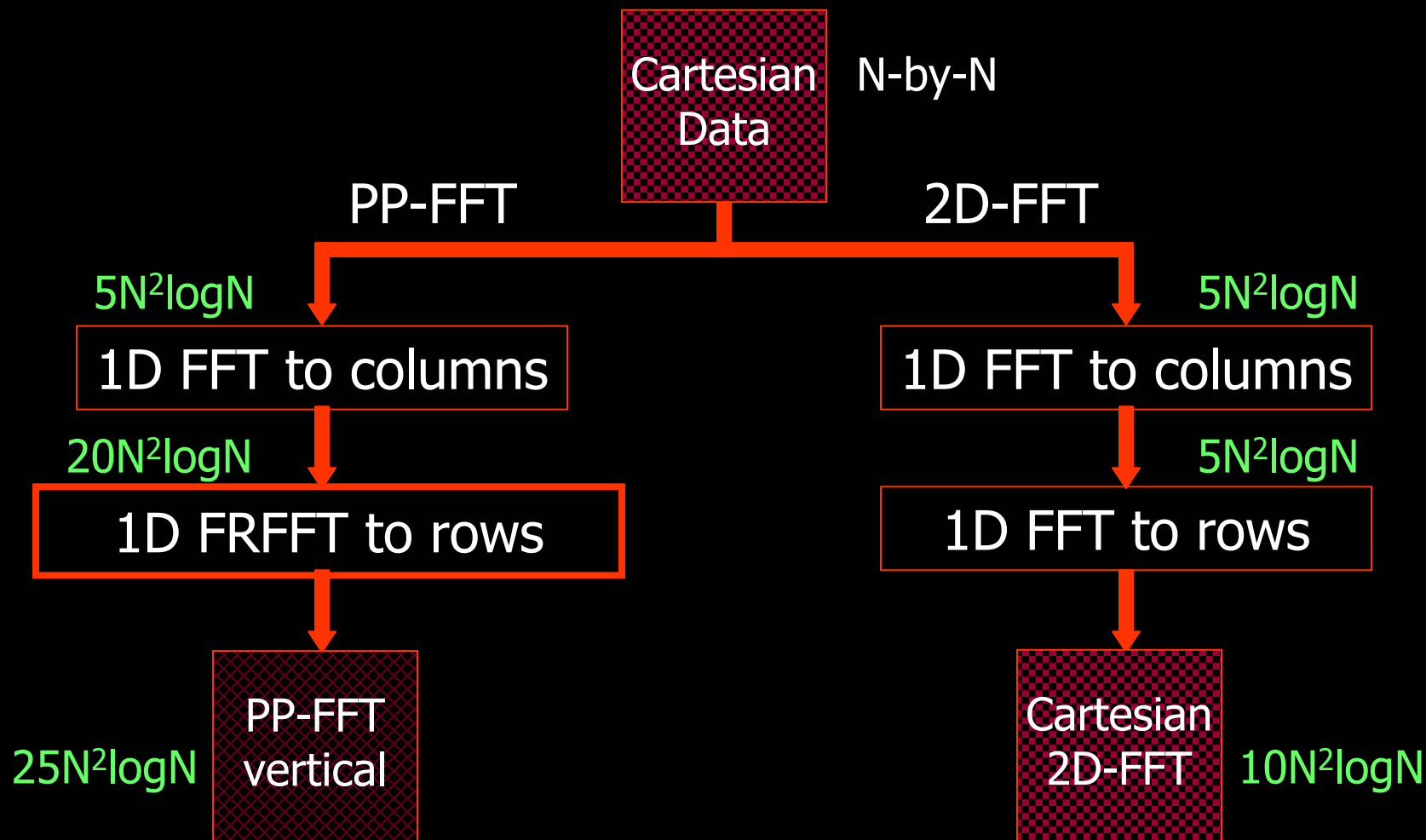
Fast Fourier Transform

- ❑ The destination samples are uniformly sampled vertically,
- ❑ Per each row, destination samples are uniformly sampled horizontally,
- ❑ Fractional Fourier Transform is the answer (Chirp-Z), with complexity: $O(20N\log_2N)$.

[Why?]

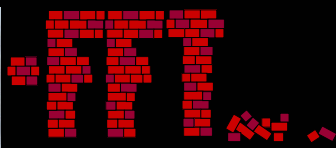


PP-FFT versus 2D-FFT




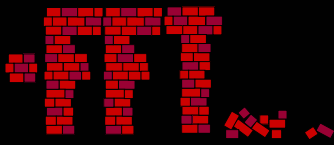
The PP-FFT - Properties

- ❑ Exact in exact arithmetic.
- ❑ No parameters involved !!
- ❑ Complexity - $O(50 \cdot N^2 \log_2 N)$ versus $O(N^4)$.
- ❑ 1D operations only.
- ❑ For the chosen grid ($S_r = S_t = 2$) - $\kappa \approx 5$.

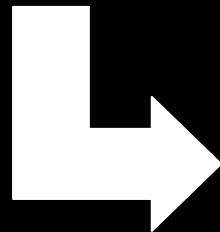
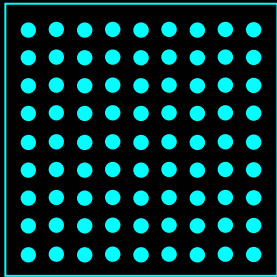


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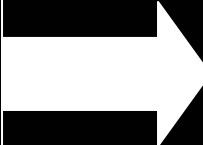
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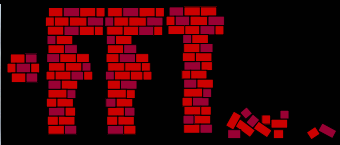
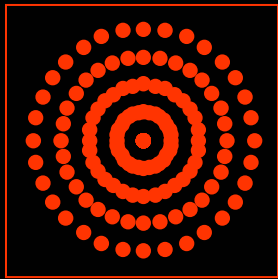
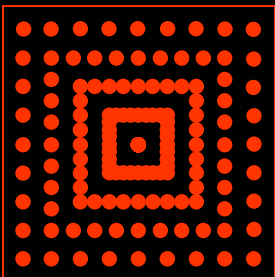
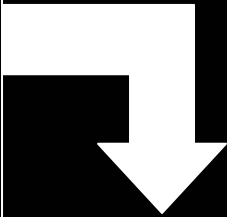
6. From Pseudo-Polar to Polar



Fast and Exact
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on a polar-like
grid

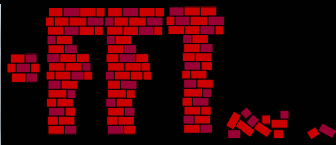
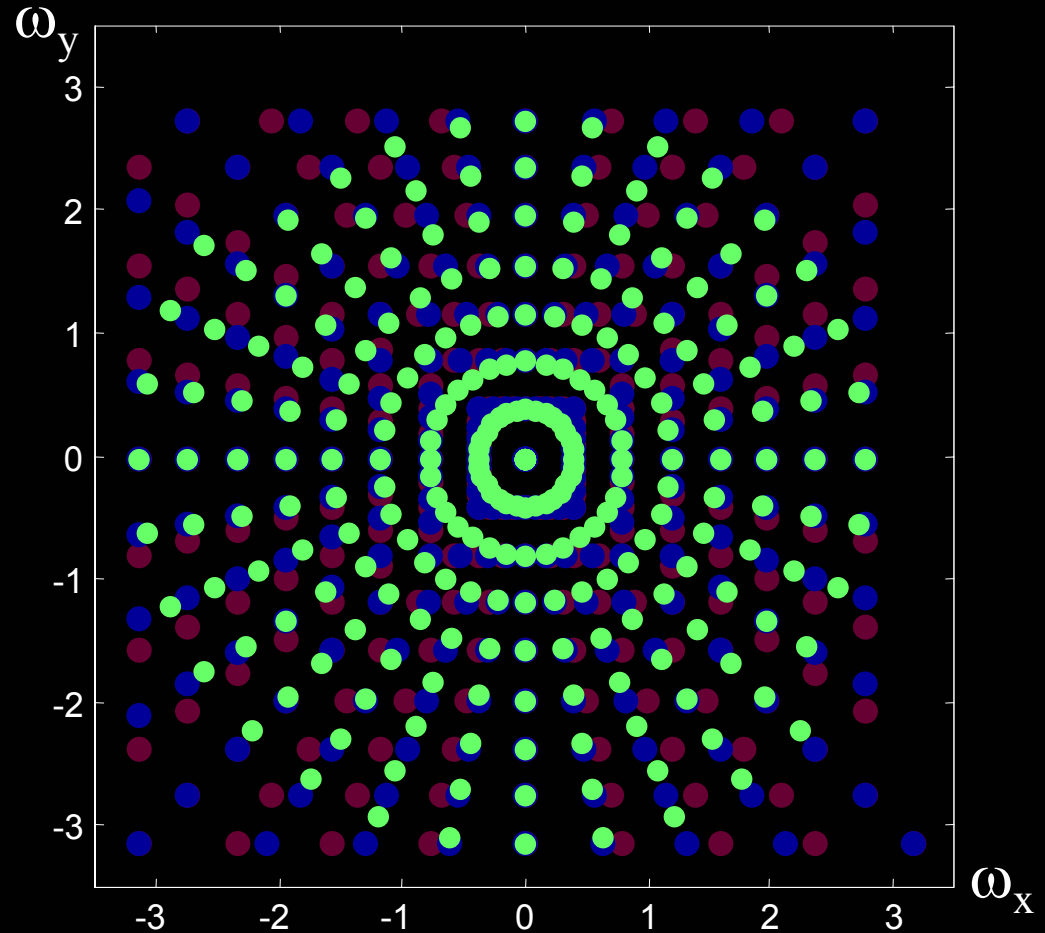


2 stages of 1D
interpolations
to get to the
polar grid



The Interpolation Stages

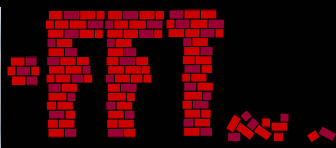
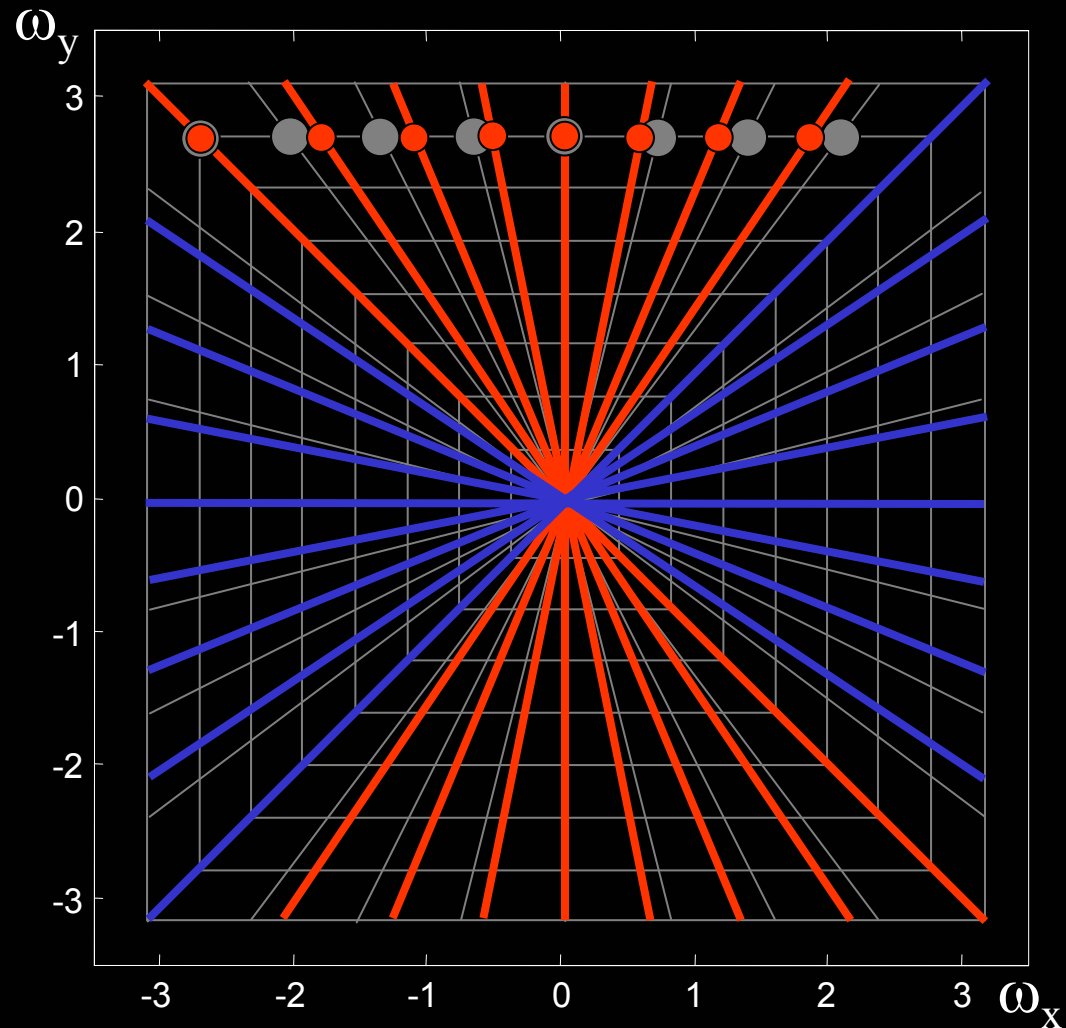
- The original Pseudo-Polar Grid
- Warping to equi-spaced angles
- Warping each ray to have the same step



First Interpolation Stage

Rotation of the rays to have equi-spaced angles (S-Pseudo-Polar grid):

- Every row is a trigonometric polynomial of order N ,
- FRFT on over-sampled array and 1D interpolation,
- Very accurate.



The Required Warping

Basically vertical lines:

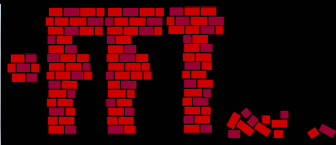
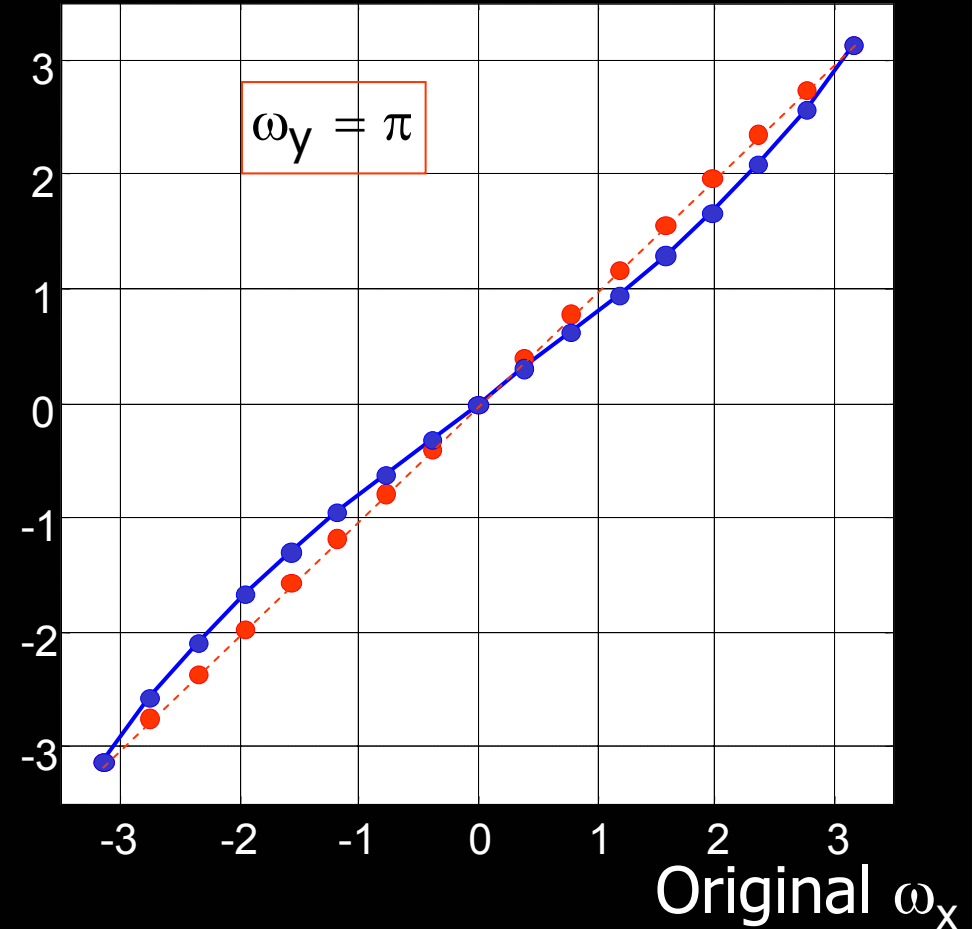
$$\left\{ \omega_y = \frac{2\pi l}{NS_r}, \omega_x = \frac{2m}{NS_t} \omega_y \right\}_{l, m = -NS_t/2}^{NS_t/2 - 1}$$



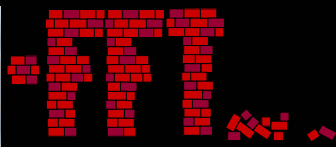
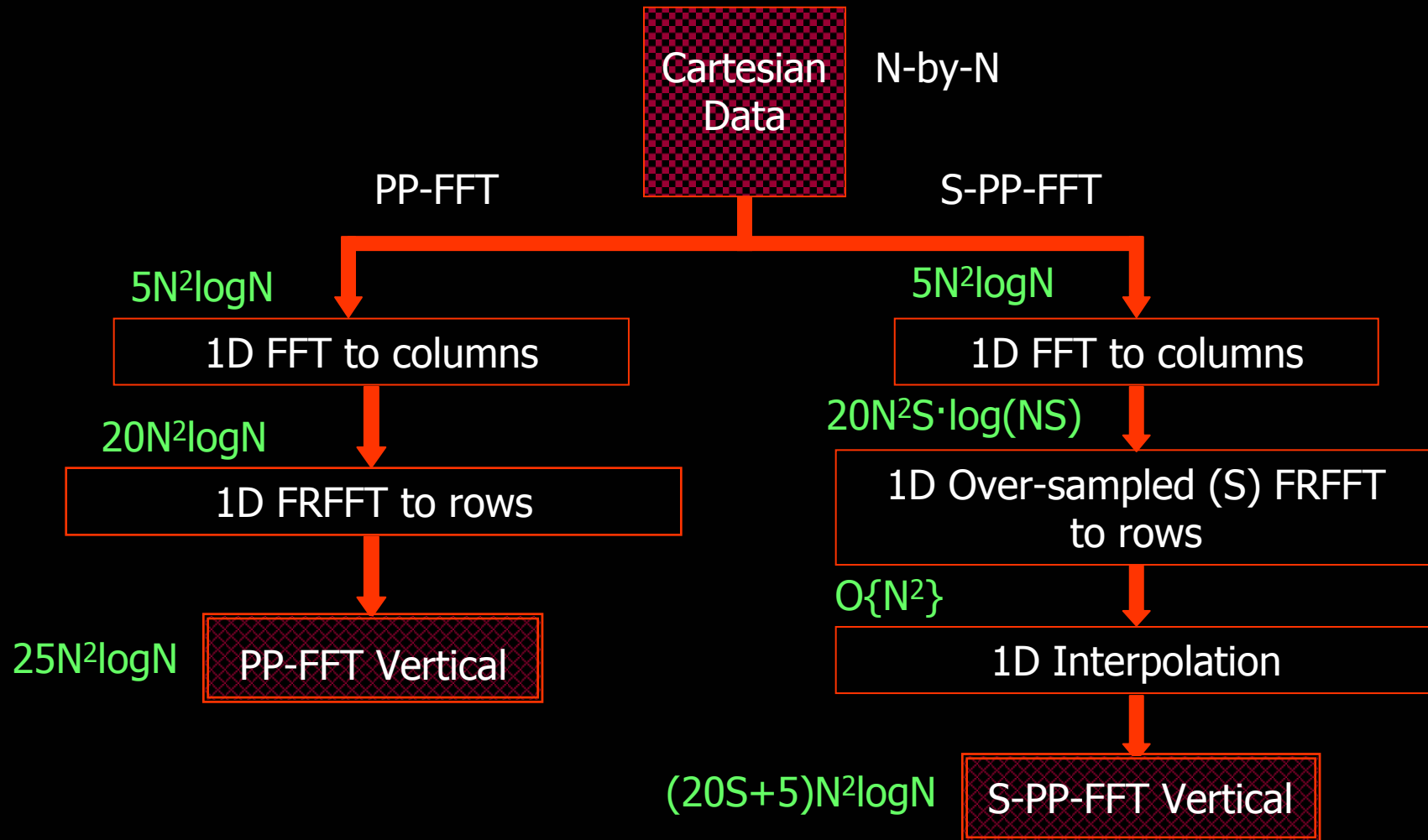
$$\left\{ \omega_x = \omega_y \cdot \tan\left(\frac{m\pi}{2NS_t}\right) \right\}_{m = -NS_t/2}^{NS_t/2 - 1}$$

[Why?]

New ω_x

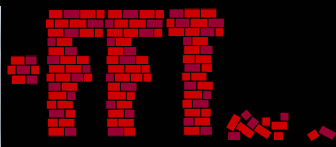
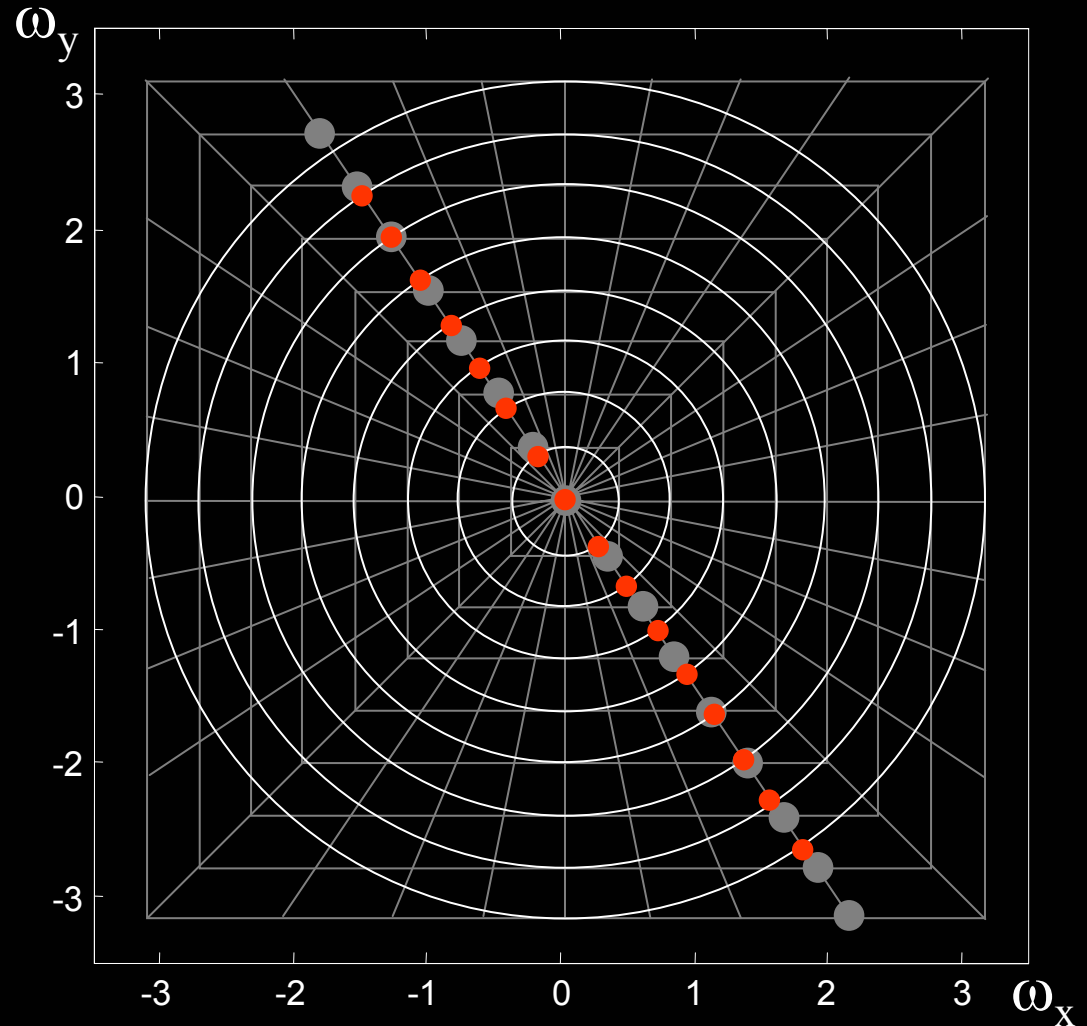


The Actual Interpolation



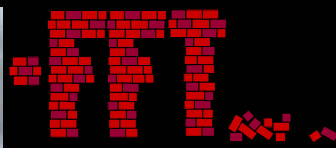
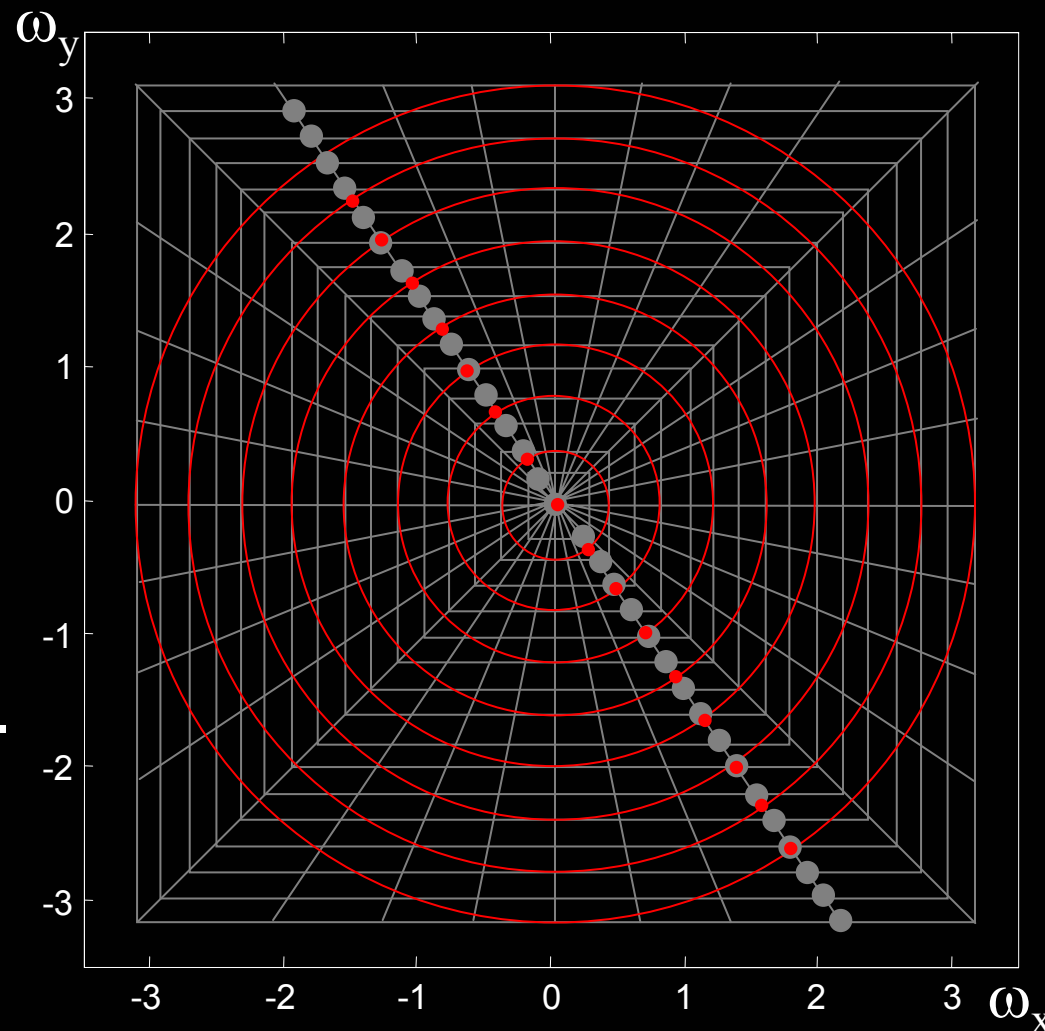
Second Interpolation Stage

- As opposed to the previous step, the rays are not trigonometric polynomials of order N ,
- We proved that the rays are band-limited (smooth) functions,
- Over-sampling and interpolation is expected to perform well.

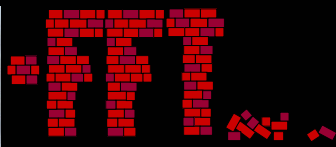
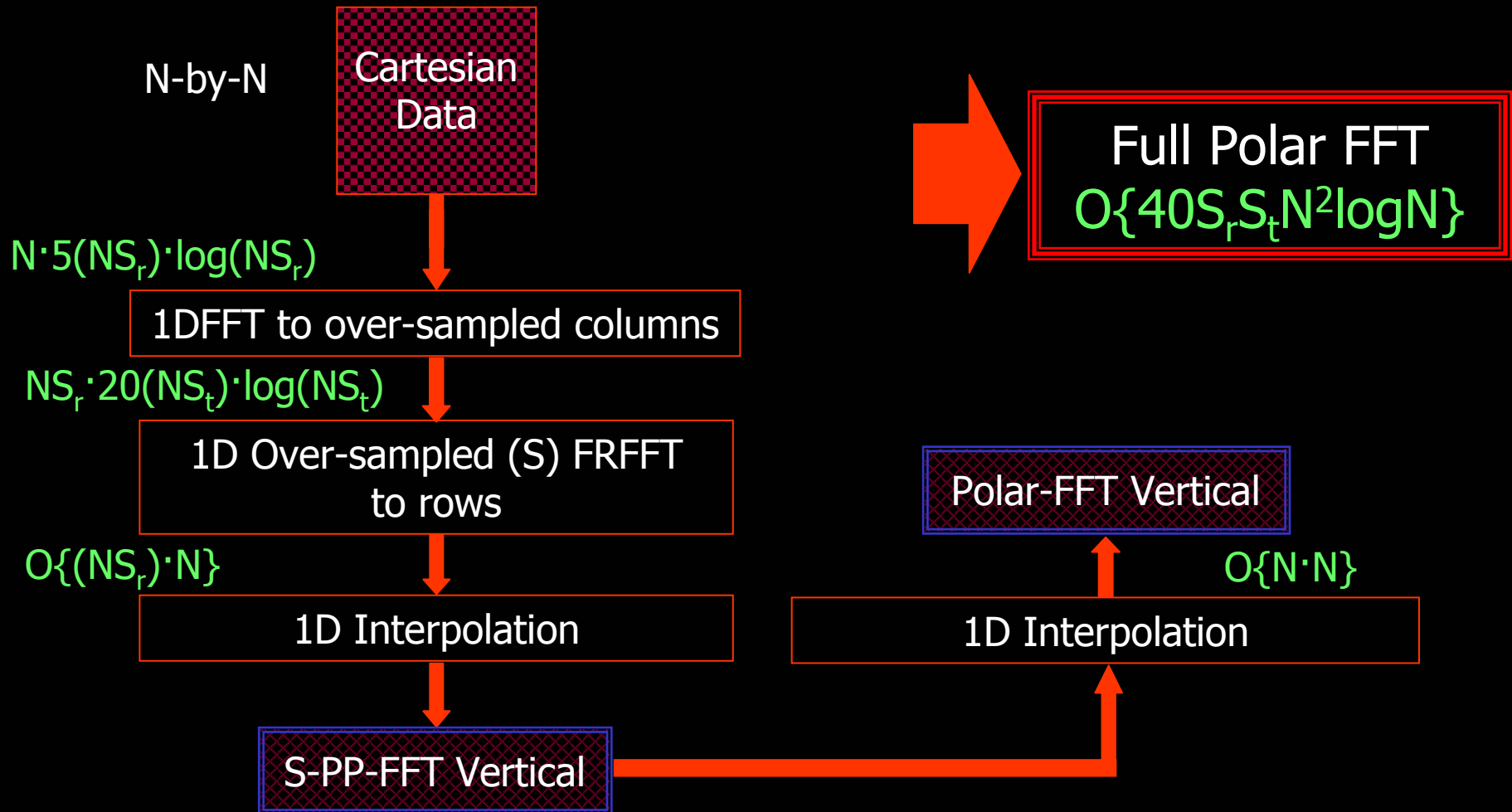


Over-Sampling Along Rays

- ❑ Over-sampling along rays cannot be done by taking the 1D ray and over-sampling it.
- ❑ $S_r > 1$:
 - More concentric squares.
 - S_r longer 1D-FFT's at the beginning of the algorithm.
 - S_r times FRFFT operations.



The Actual Interpolation



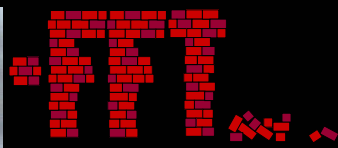
To Summarize

We propose a


Fast Polar Fourier Transform

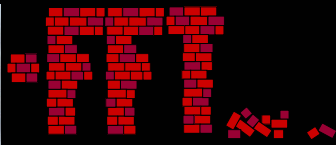
with the following features:

- Low complexity – $O(\text{Const} \cdot N^2 \log_2 N)$
- Vectorizability – 1D operations only
- Non-Expansiveness – Factor 2 (or 4) only
- Stability – via Regularization
- Accuracy – 2 orders of magnitude over USFFT methods



Agenda

1. Thinking Polar – Continuum
2. Thinking Polar – Discrete
3. Current State-Of-The-Art
4. Our Approach - General
5. The Pseudo-Polar Fast Transform
6. From Pseudo-Polar to Polar
7.  **Algorithm Analysis**
8. Conclusions



7. Algorithm Analysis

We have a code performing the Polar-FFT in Matlab:

```
Y=Polar_FFT(X);
```

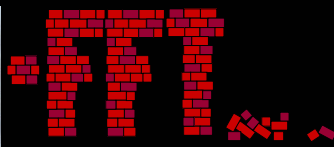
OR

```
Y=Polar_FFT(X, St, Sr);
```

Where: X – Input array of N-by-N samples

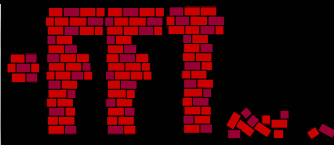
S_t, S_r – Over-sampling factors in the approximations

Y – Polar-FFT result as an 2N-by-2N array with rows being the rays and columns being the concentric circles.



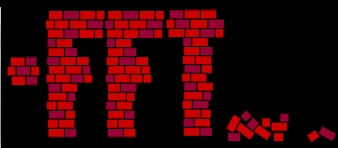
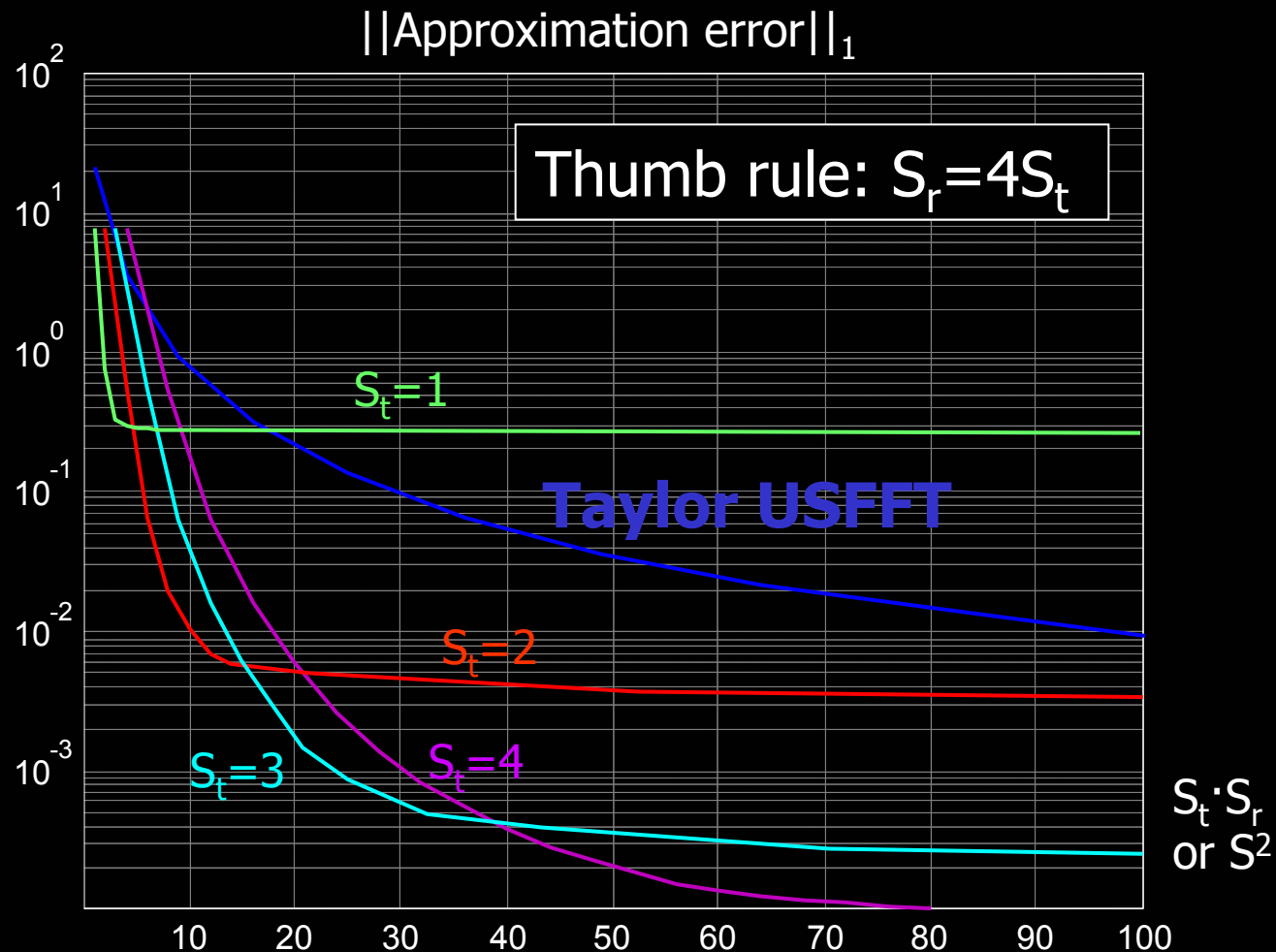
The Implementation

- ❑ The current Polar-FFT code implements Taylor method for the first interpolation stage and spline ONLY (no-derivatives) for the second stage.
- ❑ For comparison, we demonstrate the performance of the USFFT method with over-sampling S and interpolation based on the Taylor interpolation (found to be the best).



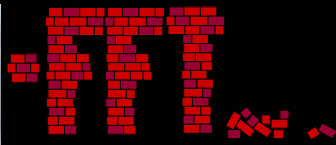
Error for Specific Signal

- Input is random 32-by-32 array,
- USFFT method uses one parameter whereas there are two for the up-sampling in the Polar-FFT.
- Thumb rule: $S_r \cdot S_t = S^2$.



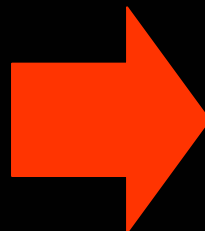
Error For Specific Signals

- ❑ Similar curves obtained of $\|*\|_{\infty}$ and $\|*\|_2$ norms.
- ❑ Similar behavior is found for other signals.
- ❑ Conclusion: For the proper choice of S_t and S_r , we get 2-orders-of-magnitude improvement in the accuracy comparing to the best USFFT method.
- ❑ Further improvement should be achieved for Taylor interpolation in the second stage.
- ❑ US-FFT takes longer due the 2D operations.

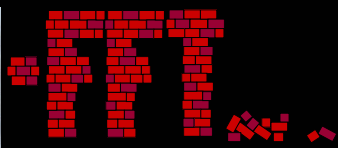


The Transform as a Matrix

All the involved transformations (accurate and approximate) are linear - they can be represented as a matrix of size $4N^2$ -by- N^2 .



$$\begin{array}{c} \text{Approximate} \\ \downarrow \\ \underline{Y}_a = \underline{A} \underline{X} \\ \text{Or} \\ \underline{Y}_t = \underline{T} \underline{X} \\ \uparrow \\ \text{True} \end{array}$$

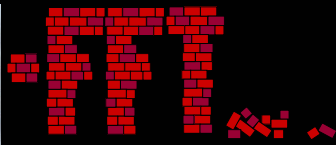


Regularization of T/A

- An input signal of N-by-N is transformed to an array or 2N-by-2N.
- For N=16, \mathbf{T} size is 1024-by-256, and $\kappa \approx 60,000$ (bad for inversion).
- Adding the assumption that the Frequency corners should be zeroed, we obtain

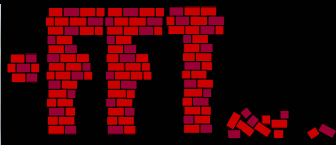
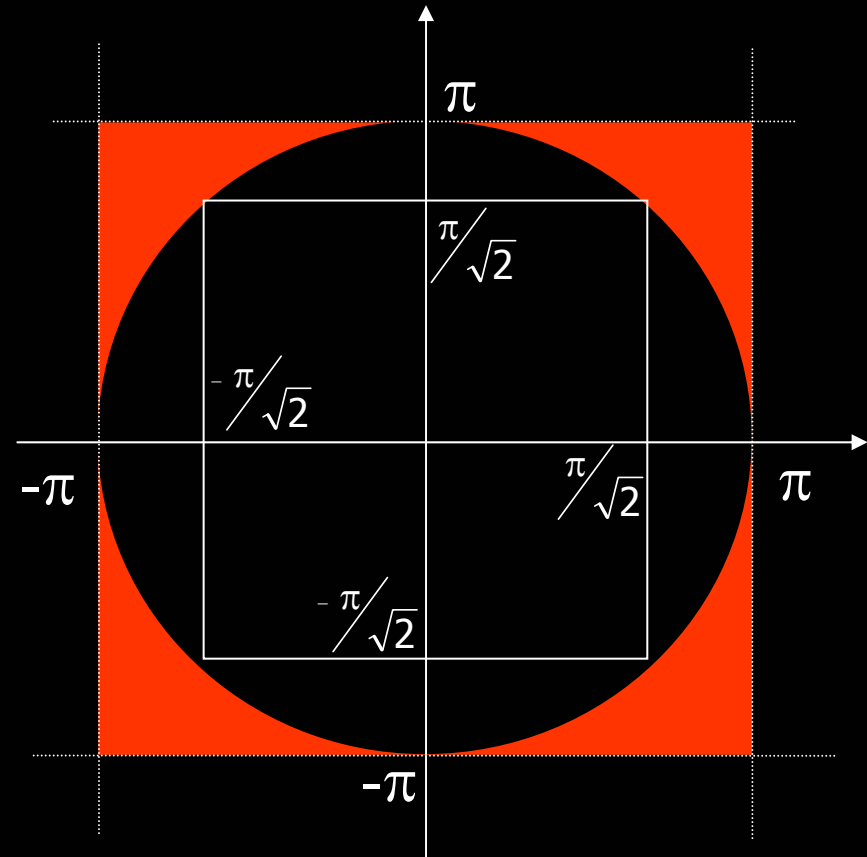
$$\underline{y} = \mathbf{T}_{\text{Polar}} \underline{x} \quad \longrightarrow \quad \begin{bmatrix} \mathbf{T}_{\text{Polar}} \\ \mathbf{T}_{\text{Corner}} \end{bmatrix} \underline{x} = \begin{bmatrix} \underline{y} \\ \underline{0} \end{bmatrix}$$

and the condition number becomes $\kappa \approx 5$!!!



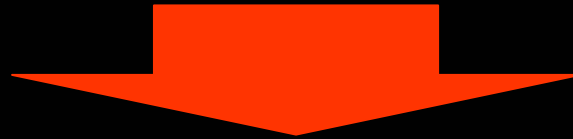
Discarding the Corners?

- ❑ If the given signal was sampled at 1.4142 the Nyquist Rate, the corners can be removed.
- ❑ If it is not, over-sampling it can be done by FFT.



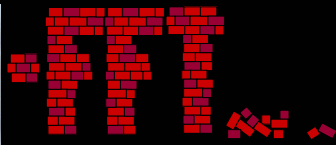
Error Analysis – Worst Signal

Approximation error is : $(\mathbf{A}_{\text{Polar-FFT}} - \mathbf{T})\underline{x} = \underline{e}_{\text{Polar-FFT}}$



$$\text{Worst error : } \{\underline{x}_{\text{worst}}, e_{\text{worst}}^2\} = \underset{\underline{x}}{\text{Arg/Max}} \frac{\|(\mathbf{A}_{\text{Polar-FFT}} - \mathbf{T}_{\text{Polar}})\underline{x}\|_2^2}{\|\underline{x}\|_2^2}$$

$$\text{Worst relative error : } \{\underline{x}_{\text{rworst}}, e_{\text{rworst}}^2\} = \underset{\underline{x}}{\text{Arg/Max}} \frac{\|(\mathbf{A}_{\text{Polar-FFT}} - \mathbf{T}_{\text{Polar}})\underline{x}\|_2^2}{\|\mathbf{T}_{\text{Polar}}\underline{x}\|_2^2}$$

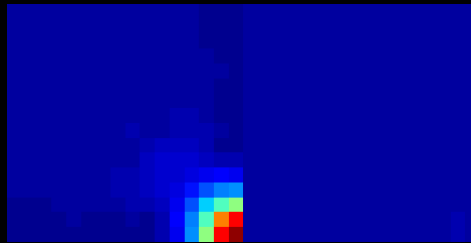


Worst Signal - Results

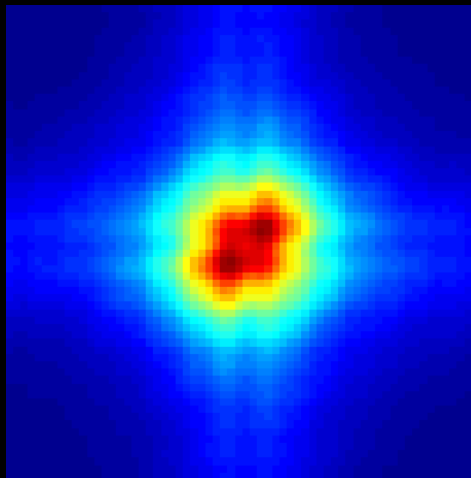
$$N=16 \rightarrow \mathbf{T} \in \mathbb{C}^{1024 \times 256}, S=S_r=S_t=4$$

USFFT

worst signal (abs. Value) $\lambda=3.469$



The worst case signal in the freq. Domain (abs. and shifted)

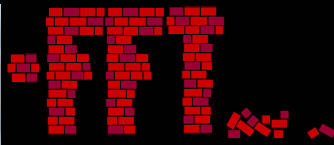
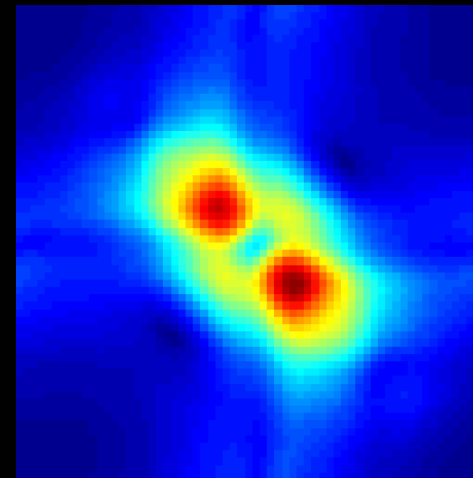


Polar-FFT

worst signal (abs. Value) $\lambda=0.0319$



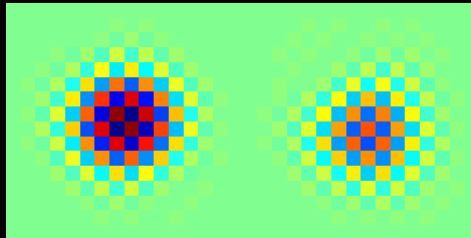
The worst case signal in the freq. Domain (abs. and shifted)



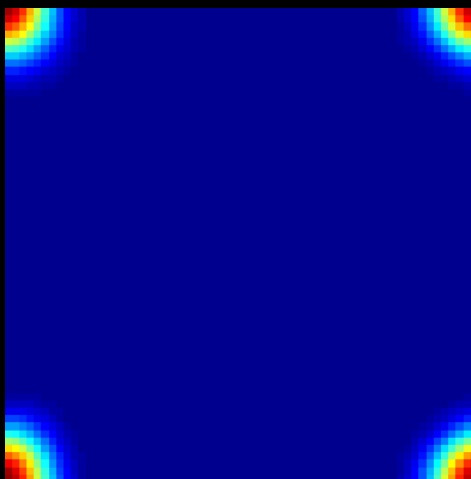
Relative Worst Signal - Results

Same parameters: $N=16 \rightarrow \mathbf{T} \in \mathbb{C}^{1024 \times 256}$, $S=S_r=S_t=4$

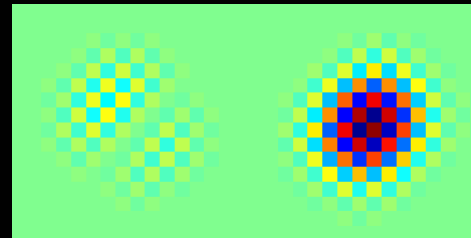
USFFT
worst signal (abs.
Value) $\lambda=0.0613$



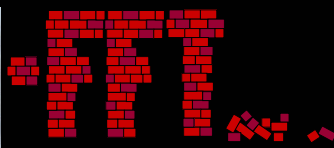
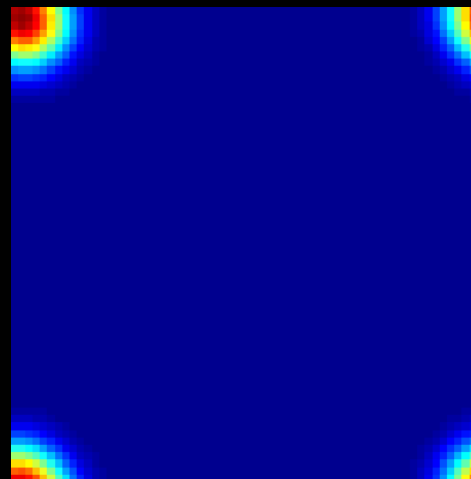
The worst case
signal in the freq.
Domain (abs. and
shifted)



Polar-FFT
worst signal (abs.
Value) $\lambda=0.0023$



The worst case
signal in the freq.
Domain (abs. and
shifted)



Comparing Approximations

- Solve for the eigenvalue/vectors of the matrix

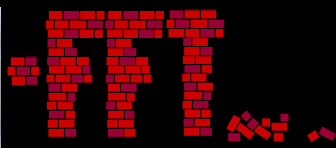
$$(\mathbf{A}_{\text{Polar-FFT}} - \mathbf{T}_{\text{Polar}})^H (\mathbf{A}_{\text{Polar-FFT}} - \mathbf{T}_{\text{Polar}})$$

and obtained $\{\lambda_k, \underline{x}_k\}_{k=1}^{N^2}$ (λ_k in ascending order).

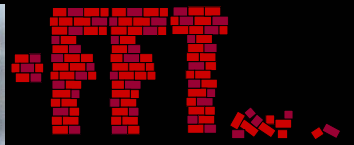
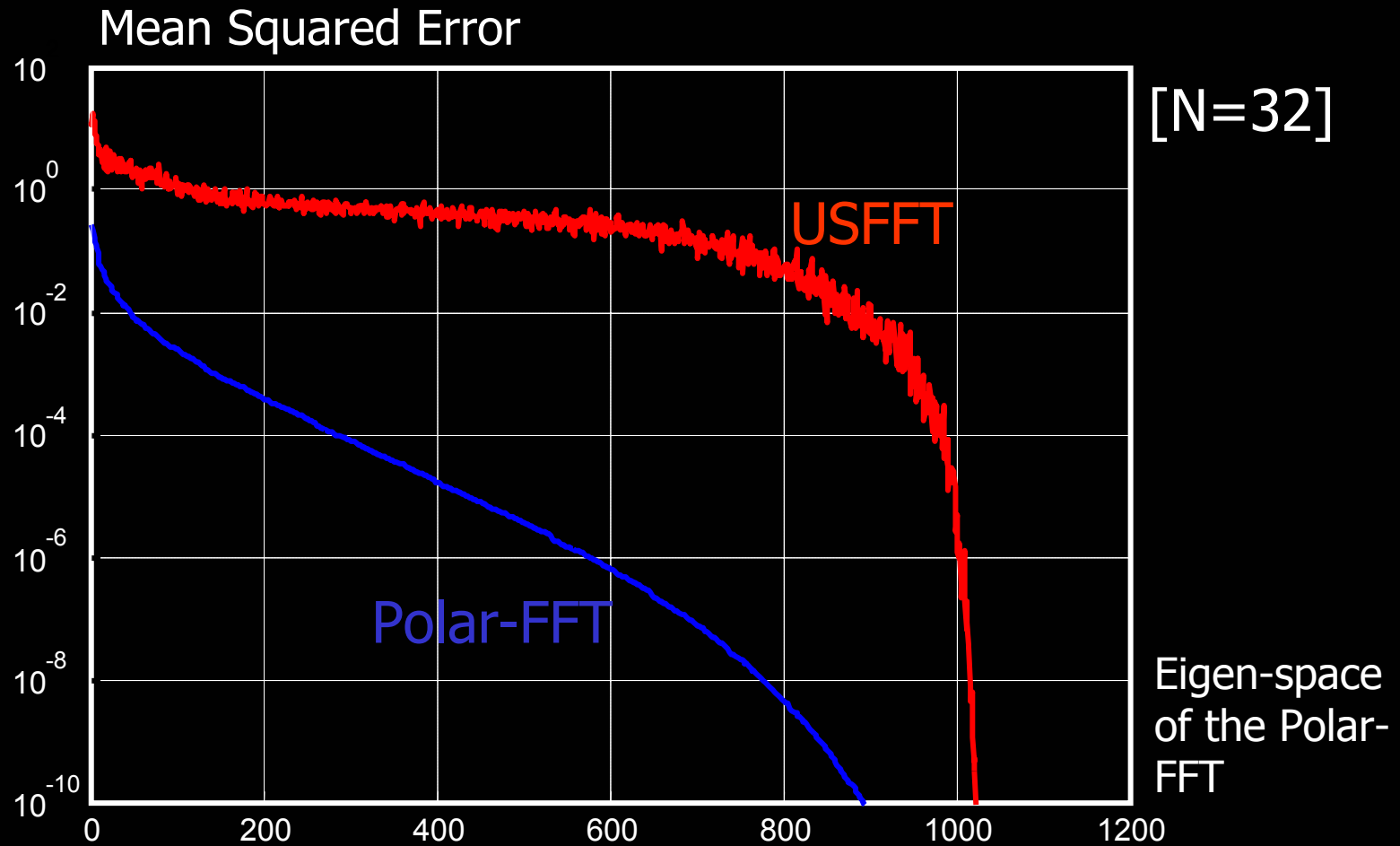
- Compare to $\mathbf{A}_{\text{USFFT}}$ by computing

$$\alpha_k = \|(\mathbf{A}_{\text{USFFT}} - \mathbf{T}_{\text{Polar}})\underline{x}_k\|_2^2$$


using the above eigenvectors and compare to λ_k .

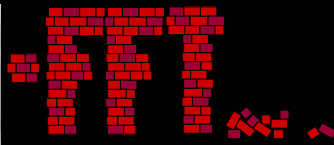


Comparing Approximations - Results



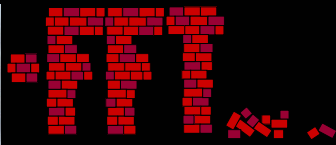
Agenda

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8.  **Conclusions**

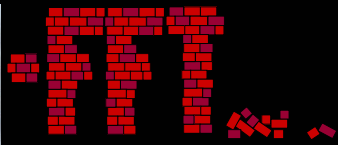


8. Conclusions

- ❑ We have proposed a fast, accurate, stable, and reliable Polar Discrete-Fourier-Transform.
- ❑ By this we extend utility of FFT algorithms to new class of settings in image processing.
- ❑ Future plans:
 - Extend the analysis and improve further,
 - Demonstrate applications,
 - Publish the code for the procedure and some applications over the internet.

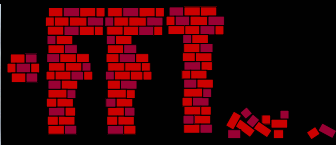


Beyond this slides are
the appendix or
redundant slides



USFFT for T^+

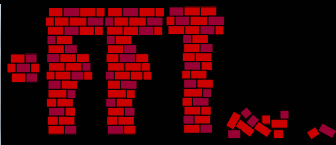
- ❑ **Over-sample** Polar grid (and possibly partial derivatives).
- ❑ Associate polar neighbors to each Cartesian grid point.
- ❑ Approximate **interpolation** to get the Cartesian grid values.
- ❑ Perform the Cartesian 2D Inverse-FFT.



Our Reading of Literature

Direct Fourier method with over-sampling and interpolation (also called gridding) – see

- ❑ Natterer (1985).
- ❑ Jackson, Meyer, Nishimura and Macovski (1991).
- ❑ Schomberg and Timmer (1995).
- ❑ Choi and Munson (1998).



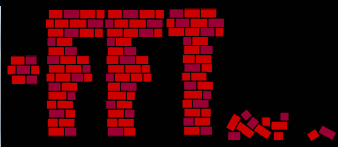
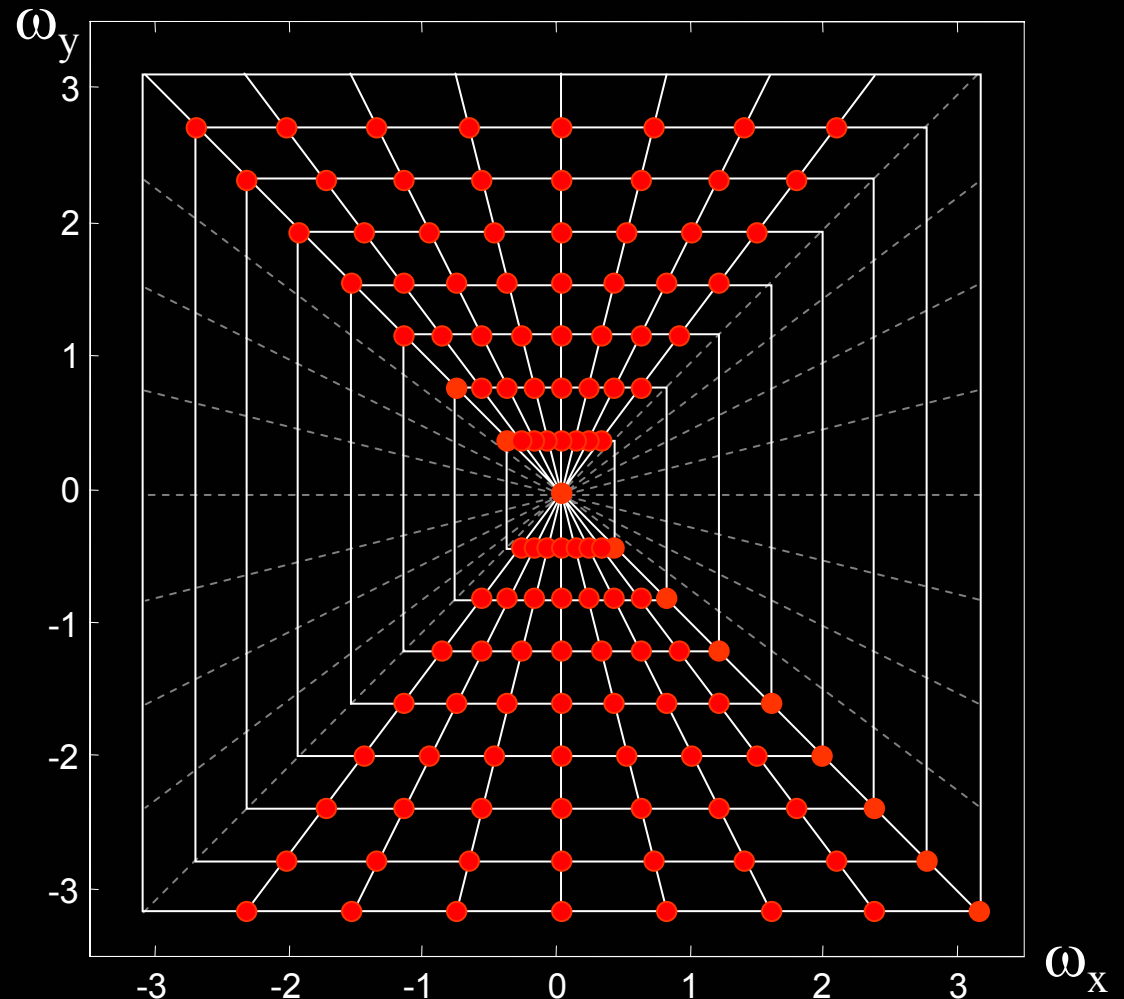
The Pseudo-Polar Sampling

Basically vertical lines:

$$\left\{ \omega_y = \frac{2\pi l}{NS_r} \right\}_{l=-NS_r/2}^{NS_r/2-1}$$

$$\left\{ \omega_x = \frac{2m}{NS_t} \omega_y \right\}_{m=-NS_t/2}^{NS_t/2-1}$$

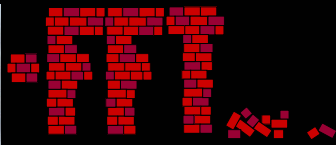
For $S_t=S_r=1$, we have
 N^2 grid points



The Pseudo-Polar FT – Stage 1

$$\begin{aligned}
 F(\omega_x, \omega_y) &= \sum_{k_1=0}^{N-1} \sum_{k_2=0}^{N-1} f[k_1, k_2] \exp\{-ik_1 \omega_x - ik_2 \omega_y\} = \\
 &= \sum_{k_1=0}^{N-1} \sum_{k_2=0}^{N-1} f[k_1, k_2] \exp\left\{-ik_1 \frac{2m}{NS_t} \omega_y - ik_2 \omega_y\right\} = \\
 &= \sum_{k_1=0}^{N-1} \exp\left\{-ik_1 \frac{2m}{NS_t} \omega_y\right\} \underbrace{\sum_{k_2=0}^{N-1} f[k_1, k_2] \exp\{-ik_2 \omega_y\}}_{=\hat{f}[k_1, \ell]}
 \end{aligned}$$

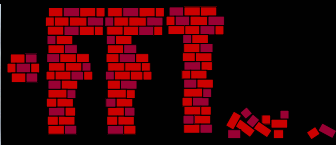
 This part is obtained by 1D-FFT along the rows !!



The Pseudo-Polar FT – Stage 2

$$F(\omega_x, \omega_y) = F[m, \ell] = \sum_{k_1=0}^{N-1} \hat{f}[k_1, \ell] \exp\left\{-ik_1 m \frac{2\omega_y}{NS_t}\right\}$$

- This summation takes columns of $\hat{f}[k_1, \ell]$ (being equi-spaced 1D signals) and computes Fourier transform of it.
- The destination grid points are also 1D equi-spaced in the frequency domain, BUT THEY ARE NOT IN THE RANGE $[-\pi, \pi]$, but rather $[-\omega_y, \omega_y]$.
- This task is called Fractional Fourier/Chirp-Z Transform.

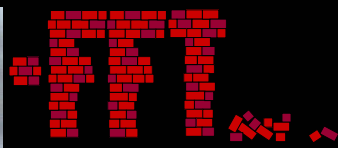


Fractional Fourier Transform

$$F[m] = \sum_{k=0}^{N-1} f[k] \exp \left\{ -i \frac{2\pi km}{N} \cdot \alpha \right\}$$

- For $\alpha=1$ we get the ordinary 1D-FFT,
- For $\alpha=-1$ we get the ordinary 1D-IFFT,
- There exists a Fast Fractional Fourier Transform with the complexity of $O(20 \cdot N \log_2 N)$, based on 1D-FFT operations.

See: Fast fractional Fourier transforms and applications, by D. H. Bailey and P. N. Swarztrauber, *SIAM Review*, 1991, and also Bluestein, Rabiner, and Rader (1960's).

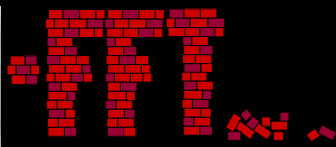


FR-FFT Detailed

$$\begin{aligned}
 F[m] &= \sum_{k=0}^{N-1} f[k] \exp\left\{-i \frac{2\pi km}{N} \cdot \alpha\right\} = \\
 &= \sum_{k=0}^{N-1} f[k] \exp\left\{-i \frac{\pi[(k-m)^2 - k^2 - m^2]}{N} \cdot \alpha\right\} =
 \end{aligned}$$

$$= \underbrace{e^{i \frac{\pi m^2}{N} \alpha}}_{\text{Post Multiplication}} \cdot \underbrace{\sum_{k=0}^{N-1} \underbrace{f[k] \cdot e^{i \frac{\pi k^2}{N} \alpha}}_{\text{Pre-Multiplication}} \cdot \exp\left\{-i \frac{\pi(k-m)^2}{N} \alpha\right\}}_{\text{Convolution}}$$

[\[Back\]](#)



Interpolation As 1D Operation

$$\begin{aligned}
 F(\omega_x, \omega_y) &= \sum_{k_1=0}^{N-1} \sum_{k_2=0}^{N-1} f[k_1, k_2] \exp\{-ik_1\omega_x - ik_2\omega_y\} = \\
 &= \sum_{k_1=0}^{N-1} \exp\left\{-ik_1 \tan\left(\frac{m\pi}{2NS_t}\right)\omega_y\right\} \sum_{k_2=0}^{N-1} f[k_1, k_2] \exp\{-ik_2\omega_y\} = \\
 &= \sum_{k_1=0}^{N-1} \exp\left\{-ik_1 \tan\left(\frac{m\pi}{2NS_t}\right)\omega_y\right\} \hat{f}[k_1, \ell]
 \end{aligned}$$

- It is a 1D operation – But it is not the Fractional-FFT.
- Can be computed by over-sampled FRFFT and interpolation.

